

# A snapback-free and high-speed SOI LIGBT with double trenches and embedded fully NPN structure

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**Abstract:** A novel 600 V snapback-free high-speed silicon-on-insulator lateral insulated gate bipolar transistor is proposed and investigated by simulation. The proposed device features an embedded NPN structure at the anode side, and double trenches together with an N-type carrier storage (N-CS) layer at the cathode side, named DT-NPN LIGBT. The NPN structure not only acts as an electron barrier to eliminate the snapback effect in the on-state within a smaller cell pitch but also provides an extra electron extracting path during the turn-off stage to decrease the turnoff loss ( $E_{\text{off}}$ ). The double cathode trenches and N-CS layer hinder the hole from being extracted by the cathode quickly. They then enhance carrier storing effect and lead to a reduced on-state voltage drop ( $V_{\text{on}}$ ). The latch-up immunity is improved by the double cathode trenches. Hence, the DT-NPN LIGBT obtains a superior tradeoff between the  $V_{\text{on}}$  and  $E_{\text{off}}$ . Additionally, the DT-NPN LIGBT exhibits an improved blocking capability and weak dependence of breakdown voltage (BV) on the P<sup>+</sup> anode doping concentration because the NPN structure suppresses triggering the PNP transistor. The proposed LIGBT reduces the  $E_{\text{off}}$  by 55% at the same  $V_{\text{on}}$ , and improves the BV by 7.3% compared to the conventional LIGBT.

**Key words:** snapback-free; fast switching; SOI LIGBT; trench gate;  $E_{\text{off}}$

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## 1. Introduction

Lateral insulated gate bipolar transistor (LIGBT) on silicon-on-insulator (SOI) is widely applied in power integrated circuits, due to its low on-state voltage, high input impedance and high current capability<sup>[1–8]</sup>. However, LIGBTs exist the contradiction relationship between the turn-off loss ( $E_{\text{off}}$ ) and the on-state voltage ( $V_{\text{on}}$ ). The lateral injection-enhanced gate transistor (LIEGT) is proposed to improve the current capability and achieve a lower  $V_{\text{on}}$ <sup>[9]</sup>. The U-shaped channel LIGBT with dual trenches is adopted to enhance carrier stored effect and obtain a low  $V_{\text{on}}$ <sup>[10, 11]</sup>. However, the lower  $V_{\text{on}}$  usually results in a higher  $E_{\text{off}}$ . To reduce the  $E_{\text{off}}$ , the shorted-anode (SA) LIGBT provides an additional electron extracting path by introducing the N<sup>+</sup> anode region, but it brings a snapback effect<sup>[12]</sup>. The separated shorted anode (SSA) LIGBT needs a large cell pitch to solve this problem by increasing  $L_s$  (the length between N-buffer and N<sup>+</sup> anode), leading to a low area efficiency<sup>[13]</sup>. For the segmented (SEG) anode LIGBT, a very large geometry ratio of the P<sup>+</sup> and N<sup>+</sup> anode is needed to eliminate the snapback effect<sup>[14]</sup>. The multi-segment anode (MSA) adopts multi-segmented P<sup>+</sup> anode and P-buried layers in anode regions to increase the anode distribution resistance and thus eliminates the snapback effect, while its improvement on  $E_{\text{off}}$  is limited<sup>[15]</sup>. The STA LIGBT suppresses the snapback effect by forming segmented deep trenches between P<sup>+</sup> anode and N<sup>+</sup> anode, but the path between the

trenches should be small enough to eliminate the snapback effect, which will limit the improvement on  $E_{\text{off}}$  on the contrary<sup>[16]</sup>.

In this paper, a novel DT-NPN LIGBT featuring an NPN structure, double trenches and an N-CS layer is proposed to achieve a better tradeoff between the  $V_{\text{on}}$  and  $E_{\text{off}}$ . The proposed device eliminates the snapback effect within a small device dimension by adjusting the parameters of the embedded NPN structure. It achieves a low  $V_{\text{on}}$  due to the carrier storing effect enhanced by the double cathode trenches and N-CS layer. Meanwhile, the double cathode trenches enhance the latch-up ruggedness. This paper is implemented by TCAD Sentaurus Device simulation tools<sup>[17]</sup>, which includes the models of high field saturation mobility, Philips unified mobility, Auger recombination, Enormal mobility, Shockley–Read–Hall recombination, and Lackner avalanche generation.

## 2. Device structure and mechanism

Fig. 1 shows the schematic cross-sectional views of the proposed DT-NPN LIGBT, double trench segment anode LIGBT (named SEG LTIGBT), and double trench separated shorted anode LIGBT (named SSA LTIGBT). The proposed LIGBT features an NPN structure at the anode side, consisting of N<sup>+</sup> anode, P-well and N-buffer. Moreover, there are cathode trench gate (CTG), cathode blocking trench (CBT) and an N-type carrier storage (N-CS) layer at the cathode side for the three devices mentioned above. The implantations for N-CS and N-buffer layer are implemented at the same time to achieve better performance without more cost. The CTG and CBT are carried out within the same process step. The key parameters used in simulations for the three devices are shown in

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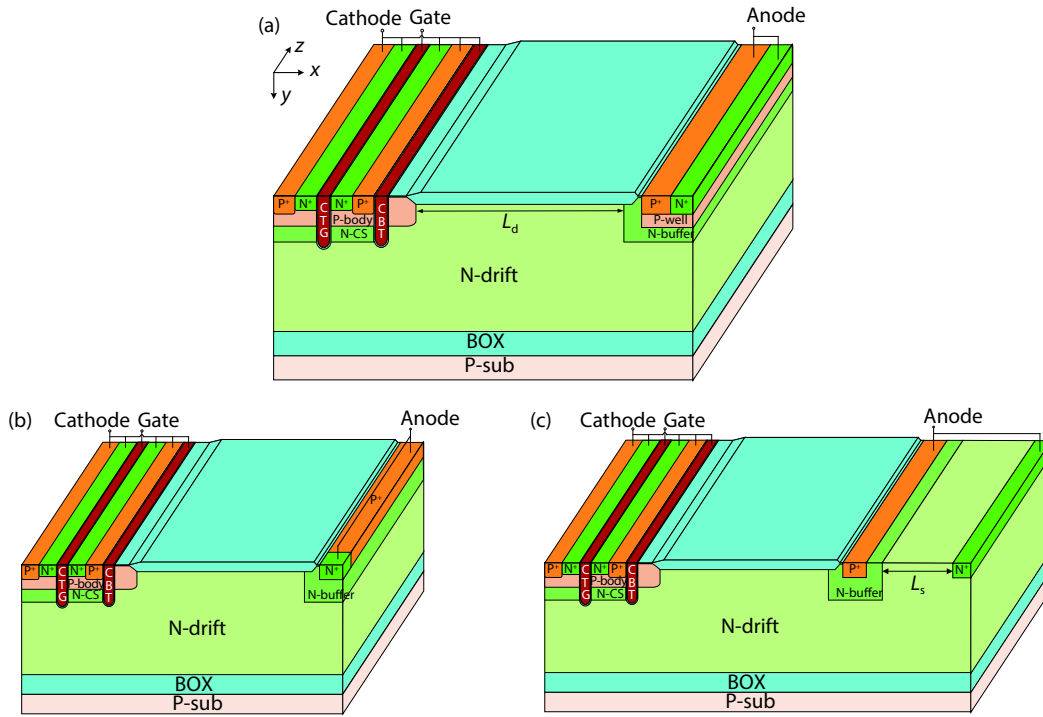


Fig. 1. (Color online) Schematic cross-sectional views of (a) DT-NPN LIGBT, (b) SEG LTIGBT, and (c) SSA LTIGBT.

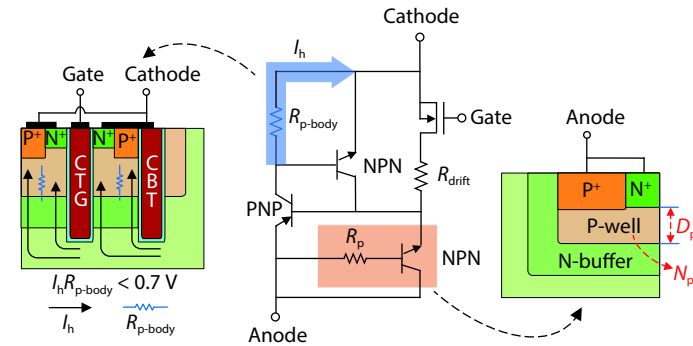


Fig. 2. (Color online) Equivalent circuit of the DT-NPN LIGBT.  $R_{p-body}$  and  $R_p$  are distributed resistance at the cathode and anode region, respectively.

Table 1. Key parameters for LIGBTs.

Parameter	DT-NPN	SEG	SSA
N-drift length, $L_d$ ( $\mu\text{m}$ )	54	54	54
SOI layer thickness, $T_s$ ( $\mu\text{m}$ )	25	25	25
Buried oxide thickness, $t_{box}$ ( $\mu\text{m}$ )	3	3	3
N-drift doping, $N_d$ ( $10^{14} \text{ cm}^{-3}$ )	2	2	2
N-buffer doping, $N_{buffer}$ ( $10^{16} \text{ cm}^{-3}$ )	6	2	6
N-CS doping, $N_{CS}$ ( $10^{16} \text{ cm}^{-3}$ )	6	6	6
P-body doping, $N_{p-body}$ ( $10^{17} \text{ cm}^{-3}$ )	1	1	1
Gate oxide thickness ( $\mu\text{m}$ )	0.1	0.1	0.1
Trench depth ( $\mu\text{m}$ )	4	4	4

Table 1.  $N_p$  and  $D_p$  are the doping concentration and depth of the P-well region at the anode side for DT-NPN LIGBT, as shown in Fig. 2. The carriers' lifetime is  $1 \mu\text{s}$  for all devices.

Fig. 2 shows the equivalent circuit of the DT-NPN LIGBT. At the initial forward conduction stage with CTG turned on, the device operates in unipolar mode and the electron current is dominant. The P-well acts as an electrons barrier and

hinders the electron from flowing through the  $N^+$  anode directly. As the anode voltage increases, the  $P^+$  anode/ $N$ -buffer junction turns on and the LIGBT quickly transforms from unipolar mode to bipolar mode without snapback effect. Meanwhile, the double cathode trenches and N-CS layer prevent holes from flowing to the cathode directly. Then, the carrier storing effect is enhanced in the N-drift and the  $V_{on}$  decreases sharply. During the turn-off stage, the shallow P-well is fully depleted and the NPN structure is activated to provide a low-resistance path to extract the electrons rapidly. Hence, the turn-off time and the  $E_{off}$  are decreased. The vertical channel and the hole bypass along the CTG and CBT respectively could decrease the distributed resistance at the cathode side, and then the proposed device enhances the immunity of latch-up effect and widens the safe operation area (SOA).

### 3. Results and discussion

Fig. 3(a) shows the dependences of BV and  $V_{on}$  on the  $P^+$  anode doping concentration ( $N_A$ ). For low  $N_A$ , the  $V_{on}$  of pro-

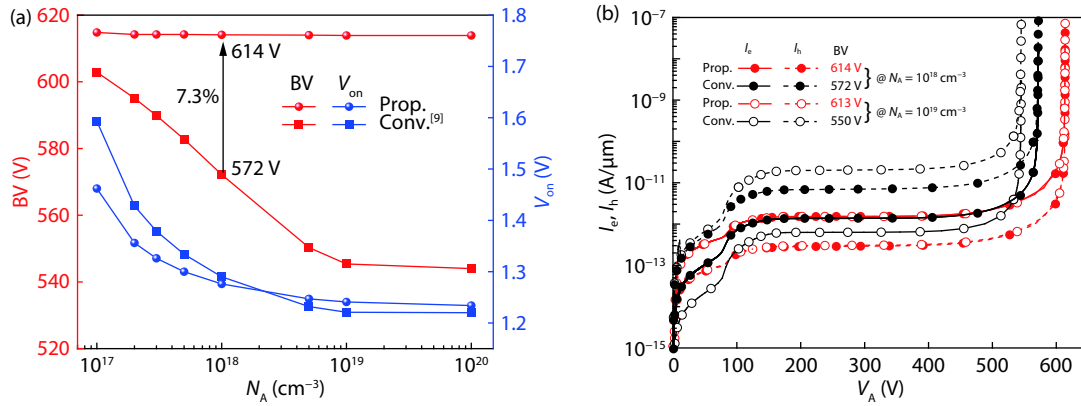


Fig. 3. (Color online) (a) Dependences of the BV and  $V_{on}$  on  $N_A$ . (b) Breakdown characteristics. Here  $I_e$  and  $I_h$  represent the electron current and hole current, respectively.

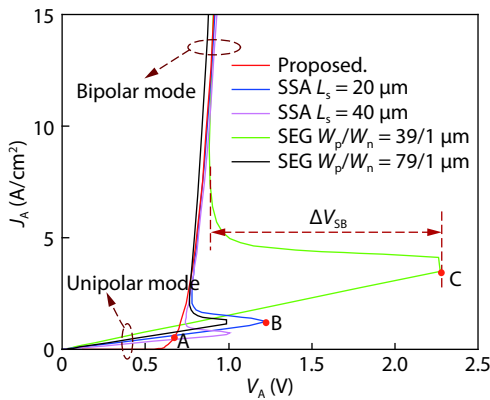


Fig. 4. (Color online) Snapback characteristics for different devices.  $W_p/W_n$  is the width of P<sup>+</sup>/N<sup>+</sup> anode.  $\Delta V_{SB}$  is snapback voltage.

posed DT-NPN LIGBT is smaller than that of the conventional (Conv.) LIGBT in Ref. [9] because the N-CS layer is dominant to increase the carrier density and decrease  $V_{on}$ . It is contrary for high  $N_A$  value, because the N<sup>+</sup> anode results in the hole injection of the proposed device much smaller than that of Conv. LIGBT. Obviously, the BV of Conv. LIGBT decreases dramatically with the increasing  $N_A$ , while it almost remains stable for the DT-NPN LIGBT for different  $N_A$ . For  $N_A = 1 \times 10^{18} \text{ cm}^{-3}$ , the BV = 614 V of the proposed LIGBT is 7.3% higher than BV = 572 V of Conv. LIGBT. Fig. 3(b) compares the current component at breakdown. The Conv. LIGBT exhibits an open-base P-N-P transistor breakdown mechanism because the leakage current increases distinctly with the increasing  $N_A$  and the hole leakage current is higher than electron leakage current. In contrast, the DT-NPN LIGBT behaves MOS-like blocking mechanism as the electron leakage current is dominated and almost irrelevant to  $N_A$ . Because the embedded NPN structure suppresses the hole injection of the P<sup>+</sup> anode/N-buffer junction. Particularly, the DT-NPN LIGBT relieves the conflict demand of  $N_A$  on high BV and low  $V_{on}$ , and then  $V_{on}$  and BV can be designed separately.

Fig. 4 illustrates the snapback characteristics of different LIGBTs with the same  $N_A$ . The proposed LIGBT eliminates the snapback effect by the optimizing the  $N_p$  and  $D_p$ , without increasing length in the  $x$ -direction. However, the SSA LTIGBT still shows a little  $\Delta V_{SB}$  even the length  $L_s = 40 \mu\text{m}$ . The SEG LTIGBT also has a snapback effect, even when the width  $W_p/W_n$  of P<sup>+</sup>/N<sup>+</sup> anode reaches 79/1  $\mu\text{m}$ . Fig. 5 shows the total

current density distribution and the flowlines extracted from Fig. 4. In Fig. 5(a), one current flowline derives from the P<sup>+</sup> anode for the DT-NPN LIGBT, which means the P<sup>+</sup> anode/N-buffer junction is turned on and the device changes from unipolar mode to bipolar mode at a smaller  $V_A$ . Owing to the barrier effect induced by the P-well, the proposed LIGBT shows snapback-free effect with compact dimension in  $x$ -direction. However, the SSA and SEG LTIGBTs require large extra dimensions above 40  $\mu\text{m}$  in  $x$ -direction and 79  $\mu\text{m}$  in  $z$ -direction respectively to reduce the  $\Delta V_{SB}$ , because all the current flowlines derive from the N<sup>+</sup> anode and both devices are still in unipolar mode, as shown in Figs. 5(b) and 5(c).

Fig. 6(a) shows the influences of  $N_p$  and  $D_p$  on the forward conduction characteristics. With the increasing  $N_p$  and  $D_p$ , the snapback effect is weakened and eventually eliminated, because larger  $N_p$  and  $D_p$  values increase the electron barrier height and width at P-well/N-buffer junction to hinder N<sup>+</sup> anode from extracting electrons, as shown in Fig. 6(b). Therefore, the increasing  $N_p$  and  $D_p$  are beneficial to decrease the  $\Delta V_{SB}$  and  $V_{on}$ , but increase the  $E_{off}$  as shown in Fig. 6(c). The snapback effect is eliminated with  $N_p = 1.5 \times 10^{16} \text{ cm}^{-3}$  and  $D_p = 1 \mu\text{m}$ , or  $N_p = 3 \times 10^{16} \text{ cm}^{-3}$  and  $D_p = 0.5 \mu\text{m}$ .

Fig. 7 compares the forward conduction characteristics of four LIGBTs with different cathode structures. The proposed LIGBT shows the lowest  $V_{on}$  of 1.25 V because both the double trenches and the N-CS layer increase the hole density at the cathode side and enhance conduction modulation effect in the drift region. The  $V_{on}$  of the LIGBT with Type1 cathode structure increases to 1.37 V, owing to the deletion of N-CS layer<sup>[18–20]</sup>. The LIGBT with Type2-cathode structure, without the CBT and N-CS layer, decreases the hole density at the cathode side and then  $V_{on}$  increases to 1.54 V. The LIGBT with a planar gate reaches the highest  $V_{on}$  of 1.94 V. The inserted hole density distribution verifies the current modulation effect induced by the N-CS, CBT and CTG, respectively.

Fig. 8 shows the switching characteristics of different LIGBTs at the same  $V_{on}$  without snapback effect. The devices start turning off at  $t_1 = 20 \mu\text{s}$ . The proposed LIGBT achieves the fastest switching speed of 54 ns because the embedded NPN structure provides a larger width of extra electron extraction path than the SEG LTIGBT and MSA LIGBT. However, the Conv. LIGBT exhibits the longest tail current and higher  $E_{off}$ , because its excess carriers stored in the drift region mainly de-

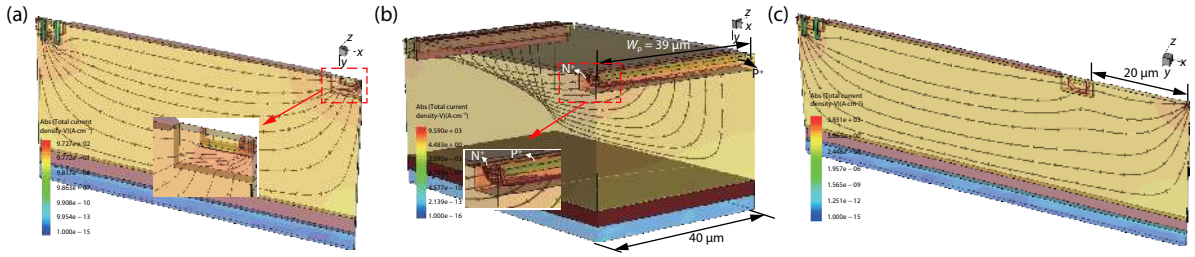


Fig. 5. (Color online) Total current density distribution and the flowlines for (a) the proposed LIGBT (at point A in Fig. 4), (b) the SEG LTIGBT (at point C in Fig. 4), and (c) the SSA LTIGBT (at point B in Fig. 4).

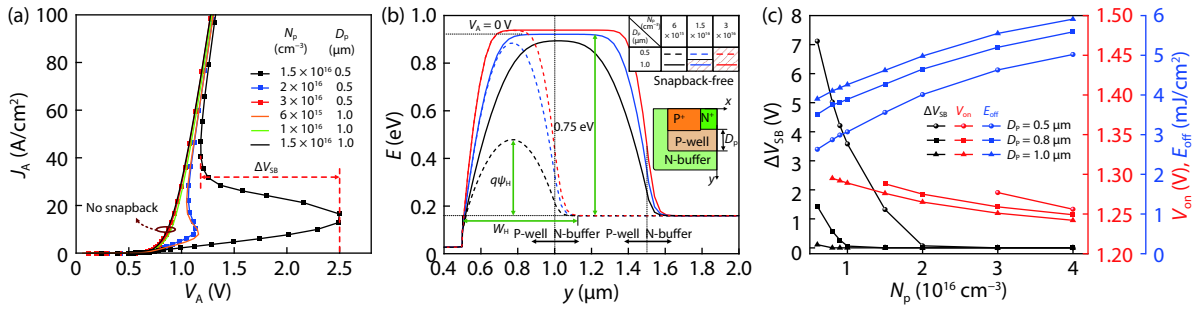


Fig. 6. (Color online) Influences of the  $N_p$  and  $D_p$  on (a) forward conduction characteristics at  $V_G = 15$  V, (b) conduction energy band distribution of P-well/N-buffer junction in the  $y$ -direction and (c)  $\Delta V_{sb}$ ,  $V_{on}$  and  $E_{off}$ .  $q\psi_H$  and  $W_H$  are the height and width of electron barrier. To entirely suppress the snapback by optimizing the  $N_p$  and  $D_p$ , the  $q\psi_H$  is about 0.75 eV.

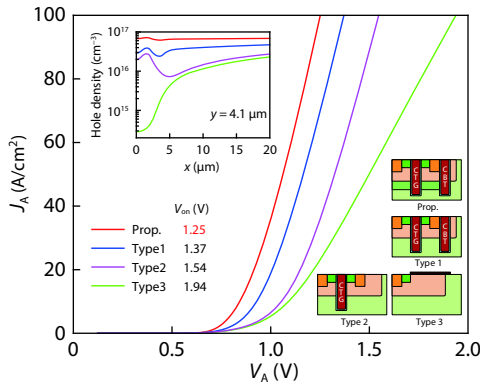


Fig. 7. (Color online) Forward conduction characteristics. Insets: hole density distribution for different LIGBTs at the cathode side (@  $y = 4.1$   $\mu\text{m}$ ) and schematic cross-sectional views of four different cathode structures for LIGBTs with the same anode structure as that of the proposed DT-NPN LIGBT.

depends on recombination to disappear without extra electron extraction path. Fig. 8(b) shows the electron concentration distributions of the proposed LIGBT and SEG LTIGBT from  $t_1$  to  $t_5$  period labeled in Fig. 8(a). It is quite obvious that the electron density of the proposed LIGBT is lower than that of the SEG LTIGBT at the same time, owing to the activated NPN with a wider rapidly electron extracting path. As shown in Fig. 8(c), a large number of current flowlines flow through the embedded NPN structure during the turn-off period of the proposed LIGBT.

Fig. 9 shows the  $I_A$ - $V_A$  characteristics for the DT-NPN LIGBT and MSA LIGBT at  $V_G = 10$  V and 15 V. In the on-state, the cathode distributed resistance  $R_{p-body}$  should be small enough to avoid triggering the parasitic NPN transistor ( $N^+$  cathode/P-body/N-CS) at the cathode side. For the proposed

LIGBT, the CBT and CTG force the hole current flowing vertically, which is conducive to decrease  $R_{p-body}$ . In contrast, the hole current inevitably flows under the  $N^+$  cathode and easily triggers the latch-up effect, even a P-buried layer is formed below the  $N^+$  cathode. Therefore, the DT-NPN LIGBT exhibits much better latch-up ruggedness than MSA LIGBT.

Fig. 10 shows the  $E_{off}$ - $V_{on}$  tradeoff of different LIGBTs at the load current density of 100 A/cm<sup>2</sup>. Due to the contribution of double trenches at cathode side and embedded NPN structure at the anode side, the proposed LIGBT shows better tradeoff than other counterparts. The  $E_{off}$  value of the proposed LIGBT is 55% lower than that of Conv. LIGBT at the same  $V_{on}$  of 1.22 V. Meanwhile, the proposed LIGBT decreases the  $V_{on}$  by 4.8%, 7.4% and 38.6% compared with SEG LTIGBT, Conv. LIGBT, and MSA LIGBT at the same  $E_{off}$ , respectively.

Fig. 11 depicts the key fabrication steps of the proposed LIGBT. In Fig. 11(a), N-CS/N-buffer implantations are implemented at the same time. To ensure the reliability of double trenches and eliminate the snapback effect, the P-body/P-well implantations are implemented under different doses and masks in Fig. 11(b). The double trenches could be formed at the same time, as shown in Fig. 11(c). All the steps can be implemented by the feasible and mature trench BCD technology.

#### 4. Conclusion

A novel 600 V class DT-NPN LIGBT is proposed and investigated. Due to the electron barrier of P-well, the proposed LIGBT entirely avoids the snapback effect within small cell pitch, while it delivers an electron extracting path during the turn-off stage to decrease  $E_{off}$ . The double cathode trenches and N-CS layer enhance the carrier storing effect to achieve a lower  $V_{on}$ . Therefore, the proposed LIGBT acquires superior

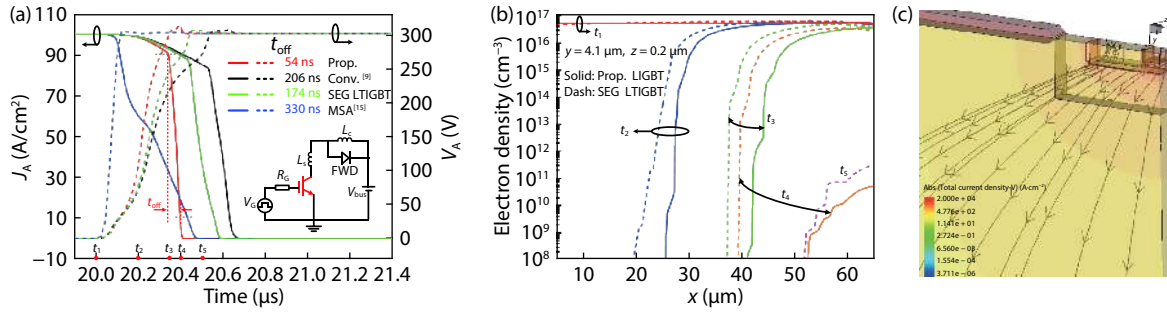


Fig. 8. (Color online) Switching characteristics: (a) switching waves, the inset shows the simulation circuit with  $R_G = 10 \Omega$  and  $L_S = 10 \text{ nH}$ , (b) carrier distribution at different time, (c) current flowlines through the embedded NPN structure at  $t_3$ .

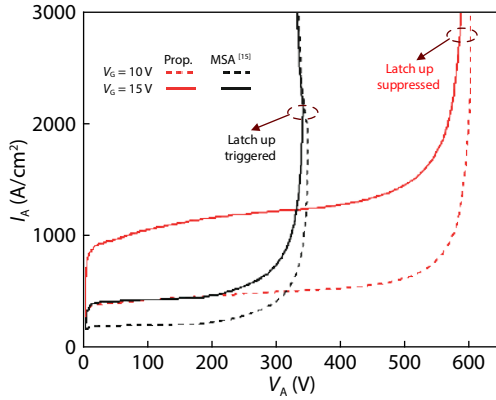


Fig. 9. (Color online)  $I_A - V_A$  characteristics for the DT-NPN and MSA LIGBT.

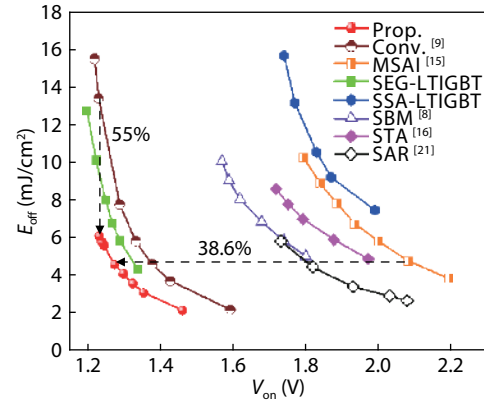


Fig. 10. (Color online)  $E_{off} - V_{on}$  tradeoff of different LIGBTs.

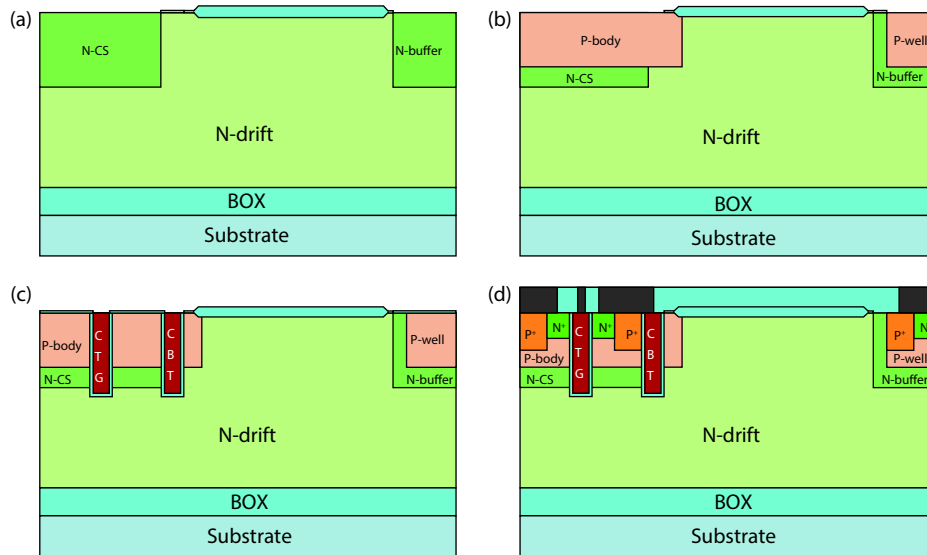


Fig. 11. (Color online) Key fabrication steps. (a) N-buffer and N-CS-layer implantations. (b) P-well and P-body implantations. (c) Trenches etching, oxidation, and poly-silicon deposition. (d) Implantations for  $N^+/P^+$  regions and formation of metal contacts.

tradeoff characteristic between  $V_{on}$  and  $E_{off}$  compared with the Conv. LIGBT, MSA LIGBT and SSA LTIGBT. The proposed LIGBT reduces the  $E_{off}$  by 55% compared to the Conv. LIGBT at the same  $V_{on}$ , and decreases the  $V_{on}$  by 38.6% compared to the MSA LIGBT at the same  $E_{off}$ . Moreover, the DT-NPN LIGBT improves the BV by 7.3% and relieves the dependence of the BV on the  $N_A$ . In addition, the DT-NPN LIGBT enhances the latch-up ruggedness, because the double trenches form vertical channels and provide a hole by-pass path to reduce the

cathode distributed resistance.

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