PHOTONICS Research

Measurement-device-independent quantum key distribution for nonstandalone networks

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Untrusted node networks initially implemented by measurement-device-independent quantum key distribution (MDI-QKD) protocol are a crucial step on the roadmap of the quantum Internet. Considering extensive QKD implementations of trusted node networks, a workable upgrading tactic of existing networks toward MDI networks needs to be explicit. Here, referring to the nonstandalone (NSA) network of 5G, we propose an NSA-MDI scheme as an evolutionary selection for existing phase-encoding BB84 networks. Our solution can upgrade the BB84 networks and terminals that employ various phase-encoding schemes to immediately support MDI without hardware changes. This cost-effective upgrade effectively promotes the deployment of MDI networks as a step of untrusted node networks while taking full advantage of existing networks. In addition, the diversified demands on security and bandwidth are satisfied, and network survivability is improved. © 2021 Chinese Laser Press

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1. INTRODUCTION

Quantum key distribution (QKD) [1–3] can share a private key securely between two authorized parties, Alice and Bob. This private key can establish unconditional secure communication combined with a one-time pad [4]. The security of QKD relies on the principles of quantum physics, with any eavesdropping on a quantum channel being detected inevitably by extra signal disturbance [5–8]. Compared with classical cryptography, the security of QKD is independent of computation complexity. Therefore, QKD is counted among the solutions to secure communication in the quantum age.

For large-scale applications, networking is imperative for QKD, which can provide secure commutation service for numerous users [9,10]. In recent years, many influential networks [11–19] have been conducted, including mature demonstrations for real-life applications. These works mark the achievement of the trusted node network and take the initial step toward the quantum Internet [20].

The trusted node network cannot provide end-to-end QKD services without the credibility of intermediary nodes. This limitation lowers the survivability of networks, which means the ability of the network to provide secure key-distribution service if there are trusted nodes is controlled by eavesdroppers. In trusted-node-based networks, if the loss of credibility happens to one node due to attacks, large parts of the network may be paralyzed (full connection is a solution but with high cost and low feasibility; Ref. [21] provides a practical and detailed analysis for the scenario). For example, an untrusted central node can deprive the star topology of function, and losing credibility of a relay node can split a line-topology network, where the star topology and the link topology are normally used to construct a quantum network [10,19] in metropolitan and wide areas, respectively. Therefore, upgrading such networks to untrusted-node-based networks for moving away from dependence on node credibility is necessary.

Any schemes that support an untrusted node can be theoretically employed in untrusted-node-based networks, such as quantum repeaters [22–25], measurement-device-independent QKD (MDI-QKD) [26–28], twin-field QKD (TF-QKD) [29–32], and device-independent QKD (DI-QKD) [33–37]. Compared with other schemes, MDI-QKD, the scheme that can completely remove all detector side-channel attacks and whose measurement unit can be regarded as an untrusted node, is the most mature and easiest one to implement [38–41] and initially shows the capability of networking [42]. Therefore, MDI-QKD is an immediate object of network upgrade [20].

Although upgrading to MDI networks can improve survivability, the cost and demand must be emphasized in network upgrades. On the cost side, the main protocol in existing networks is the BB84 protocol; however, it is incompatible with

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the MDI protocol. One major difference is the measurement mechanism. MDI-QKD requires a Bell state measurement (BSM) in the measurement unit, but BB84 does not need that [43]. Therefore, the measurement unit of the BB84 protocol cannot be used in the MDI protocol. The other difference is in the encoder of the transmitter. The bases of state preparation in QKD are X, Y, and Z, which correspond to Pauli matrices σ_x, σ_y , and σ_z , respectively. Consider the example of phase encoding in fiber-based QKD. In the BB84 protocol, all three bases can reach a low error rate, and any two of them can be used. However, in MDI protocol, only Z basis (time-bin basis) can reach a low error rate, two employed bases must contain the Z basis, and the other basis is X or Y. Although the difference in encoder can be harmonized by constraining the basis choice of BB84 protocol or postselection technique [44], the minimum requirement of the upgrade is to replace all QKD receivers, which is still a significant expenditure for device manufacturers and users. On the demand side, the requirements of two types of communication channels, control channel and data channel, in security and bandwidth are different. The control channel, which transfers command messages between devices, requires high security but low bandwidth. By contrast, the data channel requires high bandwidth but a relatively low security level. These two channel scenarios are suitable for MDI-QKD and BB84 protocols, respectively, because the former is more secure than the latter but with a lower secure key rate [45]. Therefore, considering the cost and demand, the hasty upgrade of existing networks is inadvisable, and a workable road map toward MDI networks needs to be explicit.

A viable solution to the cost and demand issues is the nonstandalone (NSA) network in 5G [46]. In the deployment of 5G, the devices of 4G also face the problem of being replaced. In addition, the few user devices that support 5G mode require a balance between the progress of the deployment and the demand. The NSA architecture is a step-by-step transition from 4G to 5G. By changing parts of devices, the 5G technology can be initially supported in 4G networks. As a result, the 5G network will be established completely as all old devices are replaced. Contrasted with the standalone (SA) network, the NSA upgrades 4G networks while taking full advantage of them, which is the most economical evolution path. Therefore, the NSA can be transplanted to QKD networks.

The central feature of the NSA is supporting a new protocol on almost-old facilities. Regardless of the NSA, the presses have reported supporting MDI on polarization-encoding SARG04 [47] and BB84 [48] and demonstrating the polarization-encoding reconfigurable network of QKD and quantum digital signature (QDS) [49]. However, the solution for phaseencoding systems is still missing, which is probably more relevant because the phase encoding has an advantage of tolerance to channel disturbance, which helps it to be widely deployed in established QKD networks [11–14,16–18] and become a mature commercialized solution. Therefore, the issues of the upgrade are imperative to phase-encoding networks, and the NSA can be harnessed in the upgrade toward MDI networks.

Here we propose an NSA-MDI scheme as the evolution of existing phase-encoding BB84 networks which harmonizes

BB84 and MDI protocols in a single system. In our design, the MDI protocol can be implemented based on existing phase-encoding BB84 networks with few hardware changes. The barrier to the deployment of MDI networks is lowered, and the maximum utilization of existing networks shows the cost-effective side of this design. In addition, the supportability of MDI and BB84 protocols not only enables the utilization of the high key rate of BB84 and the high practical security of MDI in a single network for various application scenarios but protects existing trusted-node-based BB84 networks from paralysis when the credibility of some nodes is lost. Such advantages over cost and demand benefit the manufacturers, the service providers, and the users, and then promote the deployment of MDI networks.

To achieve that, the incompatibilities of encoding and measurement between MDI and BB84 protocols must be removed. Here we redesign the optical structure of the BSM unit in MDI protocol by introducing the same structure as original BB84 systems, which is not limited to a specific realization of phase encoding. In this paper, an asymmetric Mach-Zehnder (MZ) structure [50-52] and an asymmetric Faraday-Michelson (FM) structure [18,53,54] are adopted as an example shown in the Protocol section and the Experimental System section, respectively. The new BSM of the MDI protocol is identical to the decoder of the BB84 protocol. Only a polarization controller is added to realize polarization indistinguishability for MDI protocol if it does not exist in the BB84 system using, for example, FM structures. Therefore, our evolution path can offer existing BB84 networks the capacity of MDI protocol with little hardware cost. In addition, the theoretical error rate of Xbasis can be low in our scheme (detailed theoretical proof is shown in Appendix A), which avoids the need for optical switch [55] or phase-postselection techniques [44]. To show the superiority of our evolution path, we experimentally demonstrate the BB84 protocol between Alice and Charlie, Bob and Charlie, and the MDI protocol between Alice and Bob, respectively. The experimental results are shown in the Experimental Results section. A summary is provided in the Discussion section.

2. PROTOCOL

Our MDI-QKD system is based on phase encoding and is schematically shown in Fig. 1. Structurally, an asymmetric Mach–Zehnder interferometer (AMZI) is placed on Charlie to harmonize the MDI protocol with the BB84 protocol. This additional structure helps him not only communicate with Alice and Bob as a legal user using BB84 protocol, respectively, but also play the untrusted relay of MDI-QKD between Alice and Bob. It also can reduce the theoretical error rate of the *X* basis in phase-encoding decoy-state MDI-QKD, which avoids extra payments of fast optical switch and phase-postselection techniques that the previous phase-encoding MDI-QKD schemes are needed to realize a phase shift compensation or a conversion from phase information to polarization information in Ref. [55] and reduce the error rate in the decoy-state protocols in Ref. [44], respectively.

In our system, the basis is chosen from $\mathscr{B} = \{X, Y\}$. The X basis consists of $|+\rangle = \frac{1}{\sqrt{2}}(s + |l\rangle)$ and $|-\rangle = \frac{1}{\sqrt{2}}(|s\rangle - |l\rangle)$,



Fig. 1. Schematic diagram of the nonstandalone MDI protocol. PM, phase modulator; Laser, pulsed weak-coherent source; BS, beam splitter; SPD, single-photon detector.

and the Y basis consists of $|+i\rangle = \frac{1}{\sqrt{2}}(|s\rangle + i|l\rangle)$ and $|-i\rangle = \frac{1}{\sqrt{2}}(|s\rangle - i|l\rangle)$, where $|s\rangle$ and $|l\rangle$ represent the time-bin states traveling along the short and long arm of AMZI, respectively. In addition, the three-intensity-decoy-state schemes [56,57] are adopted in both BB84 and MDI protocols. Specifically, the intensity of each laser pulse is randomly chosen from $\mathcal{I} = \{\mu, \nu, \omega\}$, and the intensities satisfy $\mu > \nu + \omega$ and $\nu > \omega \ge 0$. We use P_i^{β} to denote the probability that a laser pulse is prepared at a basis of $\beta \in \mathcal{B}$ and an intensity of $i \in \mathcal{I}$, respectively. For each instance of protocols, the intensities, \mathcal{I} , and the probabilities, P_i^{β} , are optimized for the maximum secure key rate.

A. MDI

For the MDI protocol, Alice and Bob randomly prepare their quantum state on bases β_a and β_b , respectively. Each inceptive pulse is divided into two adjacent pulses by their AMZIs. A relative phase is introduced by the phase modulator of AMZI according to the selected basis and key. The relations of the modulated phase to basis and key are listed in Table 1, where θ_a and θ_b represent the relative phase modulated by Alice and Bob, respectively. In addition, the intensities of their laser pulses are modulated as t_a and t_b , respectively, and the corresponding probabilities are $P_{t_a}^{\beta_a}$ and $P_{t_a}^{\beta_a}$, respectively.

Then Alice and Bob send their pulses to Charlie for BSM. For each pair of quantum states, unlike the original, the Charlie's AMZI further divides the incident pulses into three states of time stamps, $|ss\rangle$, $|sl(ls)\rangle$, and $|ll\rangle$. Charlie detects the middle one with two single-photon detectors (SPDs) because it contains the phase information of Alice and Bob. After basis sifting, the coincidence counting is retained, which represents a successful BSM, and other events are discarded. Here we briefly show the probabilities of valid responses, Q, on the conditions that $\theta_a = \theta_b$ and $\theta_a \neq \theta_b$ and the error rates, E,

Table 1. Code Table in MDI Protocol

	$ +\rangle$	->	$ +i\rangle$	$ -i\rangle$
θ_a	0	π	$\frac{\pi}{2}$	$\frac{3\pi}{2}$
θ_b	0	π	$\frac{\pi}{2}$	$\frac{3\pi}{2}$

$$\begin{aligned} Q_{l_{a}l_{b}}^{X}|_{\theta_{a}=\theta_{b}} &= \frac{l_{a}l_{b}}{2}, \\ Q_{l_{a}l_{b}}^{X}|_{\theta_{a}\neq\theta_{b}} &= 0, \\ E_{l_{a}l_{b}}^{X} &= \frac{Q^{X}|_{\theta_{a}\neq\theta_{b}}}{Q^{X}|_{\theta_{a}=\theta_{b}} + Q^{X}|_{\theta_{a}\neq\theta_{b}}} = 0, \\ Q_{l_{a}l_{b}}^{Y}|_{\theta_{a}=\theta_{b}} &= \frac{(l_{a}+l_{b})^{2} - 2l_{a}l_{b}}{8}, \\ Q_{l_{a}l_{b}}^{Y}|_{\theta_{a}\neq\theta_{b}} &= \frac{(l_{a}+l_{b})^{2} + 2l_{a}l_{b}}{8}, \\ E_{l_{a}l_{b}}^{Y} &= \frac{Q^{Y}|_{\theta_{a}=\theta_{b}}}{Q^{Y}|_{\theta_{a}=\theta_{b}} + Q^{Y}|_{\theta_{a}\neq\theta_{b}}} \\ &= \frac{l_{a}^{2} + l_{b}^{2}}{2(l_{a}+l_{b})^{2}} \stackrel{l_{a}=l_{b}}{=} \frac{1}{4}. \end{aligned}$$

In these equations, for simplification, the dark-count rate and afterpulse probability of the detector are neglected, the detection efficiency and transmittance are regarded as 100%, and the reference frames of Alice and Bob are aligned, which means $\theta_c = 0$. The realistic version is shown in detail in Appendix A. Note that because $Q^X|_{\theta_a\neq\theta_b} = 0$, we regard the responses under $\theta_a = \theta_b$ and $\theta_a \neq \theta_b$ as correct and error responses of the X basis, respectively. On the contrary, for the Y basis, the responses under $\theta_a \neq \theta_b$ are regarded as correct responses for a low error rate.

According to Eq. (1), the error rate of the *X* basis can be very low, which reflects the same characteristic as the *Z* basis used in the original scheme [39,40,57]. Therefore, our scheme can realize a phase-encoding MDI protocol by only modulating the phase. Neither optical switch nor the phase-postselection technique is required. Such an encoding scheme is also consistent with the phase-encoding BB84 p_{e}^{r} otocol.

Finally, with the data of $Q_{l_a l_b}^{\beta}$ and $E_{l_a l_b}^{\beta}$, the secure key can be extracted from the data when both Alice and Bob encode their bits using signal states (μ) on the X basis. The single-photon yield and error yield can be estimated by the rest of the data and engaged in the calculation of the secure key rate [57]. The SKR is given by

$$R = P_{\mu}^{X^{2}}(\mu^{2}e^{-2\mu}Y_{11}^{X,L}(1 - H_{2}(e_{11}^{Y,U})) - Q_{\mu_{a}\mu_{b}}^{X}f_{e}H_{2}((E_{\mu_{a}\mu_{b}}^{X}))),$$
(2)

where $Y_{11}^{X,L}$ is the lower bound of the yield of single-photon pairs, $e_{11}^{X,U}$ is the upper bound of the phase-flip error rate, $H_2(x) = -x \log_2(x) - (1 - x) \log_2(1 - x)$ is the binary Shannon entropy function, and f_e is the error correction efficiency. The calculation of secure key rate is also shown in detail in Appendix B.

B. BB84

For BB84 protocol, both Alice and Bob can communicate with Charlie. Here we provide a detailed description of the protocol by the example of the communication between Alice and Charlie. Alice first prepares her quantum state at a basis β_a and an intensity ι_a with probabilities $P_{\iota_a}^{\beta_a}$. The AMZI also divides each laser pulse into two adjacent pulses. The phase modulators modulate the relative phase, θ_a , between them according to the random basis and key Alice selected. The relations of the modulated phase to basis and key are listed in Table 2.

	$ +\rangle$	->	$ +i\rangle$	$ -i\rangle$
$\overline{\theta_a(\theta_b)}$	0	π	$\frac{\pi}{2}$	$\frac{3\pi}{2}$
θ_{c}	0	0	$\frac{\pi}{2}$	$\frac{\pi}{2}$

Then Alice sends her quantum state to Charlie. Charlie selects a basis β_c with probabilities P_c^{β} by modulating the relative phase, θ_c . The relation between θ_c and basis is also listed in Table 2. After the transmission along Alice's and Charlie's AMZIs, the original pulse is split into three parts according to different paths: two short arms ($|ss\rangle$), one short and one long arms ($|sl(ls)\rangle$), or two long arms ($|ll\rangle$). Charlie detects the $|sl(ls)\rangle$ with two SPDs. In all possible outcomes, no detection events are discarded, and the others are counted as valid responses if the bases of Alice and Charlie are identical. Specifically, due to the random assignment of a bit value, the double-click events cause 50% of the error rates [58]. Then, the yield, Q_t^{β} , and error rate, E_t^{β} , of a basis β and an intensity ι can be obtained from statistics.

Finally, with the data of Q_i^{β} and E_i^{β} , the parameters that are required in the calculation of the secure key rate can be estimated and bounded by decoy technology [56]. The secure key rate can be obtained by

$$R = \frac{1}{N} \left(s_0^X + s_1^X (1 - H_2(e_{1,p}^X)) - \lambda_{\rm EC} - 6\log_2 \frac{21}{e_{\rm sec}} - \log_2 \frac{2}{e_{\rm cor}} \right),$$
(3)

where s_0 is the number of vacuum events, s_1 is the number of single-photon events, $e_{1,p}$ is the phase error rate, N is the total

3. EXPERIMENTAL SYSTEM

The experimental setup is schematically introduced, as shown in Fig. 2. The laser and intensity modulator (IM) 1 of Alice and Bob compose the weak coherent pulse source. The laser is a frequency-locked continuous-wave source whose central wavelength is locked to a molecular absorption line at 1542.38 nm, with a precision of 0.0001 nm, corresponding to an approximately 10 MHz accuracy in the spectrum domain. Then, IM1 chops the continuous-wave laser into pulses with a 2.5 ns temporal width and 40 MHz repetition rate.

IM2 and an electronic variable optical attenuator (EVOA) of Alice and Bob modulate the intensity of pulses for decoystate technology [59–62]. In our system, the three-intensity decoy-state method is used, and these three intensities are denoted by μ , ν , ω , respectively. IM2 modulates the intensity according to the decoy-state method. The EVOA enables single-photon attenuation of the modulated pulses.

The asymmetric Faraday–Michelson interferometers (AFMIs) replace the AMZIs for robustness and are used for the phase-encoding quantum state preparation. The two arms of an AFMI are called the short arm (s) and the long arm (l), respectively. Specifically, for each AFMI, each laser pulse is split into two adjacent pulses by the beam splitter (BS). Then the



Fig. 2. Experimental setup for the nonstandalone MDI-QKD system. Alice and Bob can implement phase-encoding MDI-QKD and generate secure key with Charlie via BB84. Laser, frequency-locked lasers; IM1, intensity modulator as pulse generator; IM2, intensity modulator as decoy state generator; BS, beam splitter; PM, phase modulator; PS, phase shifter; FM, Faraday mirror; EVOA, electronic variable optical attenuator; EPC, electronic polarization controller; Circ, circulator; SPD, single-photon detector. For Alice, Bob, and Charlie, the combination of one BS, one phase controller, and two FMs constitutes their own AFMI; the other PM is used for phase randomization.

phase modulator (PM) modulates the relative phase between them. Here the modulated relative phases of Alice, Bob, and Charlie are denoted by θ_a , θ_b , and θ_c , respectively, and $\theta \in \{0, \pi, \pi/2, 3\pi/2\}$, which correspond to $|+\rangle, |-\rangle, |+i\rangle$, $|-i\rangle$, respectively. The Faraday mirror (FM) rotates the polarization to compensate for the disturbances caused by birefringence within the asymmetric Faraday–Michelson interferometer (FMI).

Charlie is linked to Alice and Bob by a 10 km long optical fiber, respectively, corresponding to 1.96 dB of loss for each link. Charlie's electronic polarization controller (EPC) is used to guarantee the polarization indistinguishability of Alice's and Bob's states in the MDI protocol. A dual-EPC configuration [42] or polarization scrambling method [57] is more efficient in a field environment, which is a trade-off between performance and cost. After the decoding of Charlie's FMI, the laser pulses become a superposition of three time stamps state corresponding to the paths of laser pulses. In both BB84 and MDI protocols, the middle pulses are detected for Mach-Zehnder interference and BSM, respectively, by two InGaAs/InP SPDs (Qasky WT-SPD300-LN [63]) with a detection efficiency of 25% and an averaged dark count rate of 7.5×10^{-6} per gate. Moreover, the internal transmittance of Charlie's optical components is 4.2 dB.

In electronics, all intensity modulators and phase modulators are driven by homemade digital-to-analog converters (DACs). Their modulating voltages are found by scanning the outputs of the DACs. Specifically, for the three AFMIs, the voltages of the modulated relative phases can be obtained by alternately modulating two different codes of the phase modulator while maintaining the voltages of other modulators. Scanning the voltage difference between the two coding modes and recording the counts' curves of one detector in the two coding modes, the phase difference between the curves is the differential phase to which the differential voltage corresponds. Especially when conducting the MDI protocol, the phase modulator of Charlie can be used as a phase shifter to compensate reference-frame misalignment between Alice and Bob by scanning the voltage to minimize the coincidence counts when they select different phase in the X basis or the same phase in the Y basis. The system is controlled by an FPGA module (NI PCIe-7852R). The FPGA module converts the random information of basis and key to digital signals and then sends them to the DACs for encoding. The FPGA module also implements basis sifting and data collection according to the encoding information and responses of the SPDs.

4. EXPERIMENTAL RESULTS

A. MDI-QKD

We first test the system's performance by measuring the visibility of Hong–Ou–Mandel (HOM) interference on Charlie's side. We obtain a visibility of 47.8% over 20 km of single-mode fiber, which approaches the theoretical limit of 50% for weak coherence sources.

For high performance, we optimize the parameters of our system before key distribution using particle swarm optimization (PSO) [64]. Specifically, $\mu = 0.284$, $\nu = 0.057$, $\omega = 0$,



Fig. 3. Virtual network topology and link rates of our system.

 $P_{\mu}^{X} = 0.466, \quad P_{\mu}^{Y} = 0.035, \quad P_{\nu}^{X} = 0.076, \quad P_{\nu}^{Y} = 0.293, \\ P_{\omega} = 0.130.$

Then, to achieve a higher practical security, we consider finite-sized effect and adopt the three-intensity decoy-state method. Here we apply the large deviation theory, specifically, the Chernoff bound [28,65], for the fluctuation estimation in our experiment, with a fixed failure probability of $\varepsilon = 10^{-10}$ and a total number of sifted pulse pairs $N_t = 3.97 \times 10^{11}$. Finally, we obtain the secure key rate of 1.025×10^{-5} for a transmission distance of 20 km, as shown in Fig. 3. The gains and quantum bit error rates (QBERs) of our MDI-QKD system are shown in Table 3. The method to calculate the secure key rate is shown in Appendix B.

B. BB84 QKD

Similarly, we test the system's performance by measuring the visibility of the MZ interferometer. We obtain the visibilities of 99.7% (Alice-Charlie) and 99.5% (Bob-Charlie) over 10 km of single-mode fiber which approaches the theoretical limit of 100% for weak coherence sources.

Also, we first optimize the parameters of BB84 systems. For simplicity, the parameters of the Alice-Charlie system are the same as Bob-Charlie system. Specifically, $\mu = 0.538$, $\nu = 0.063$, $\omega = 0.003$, $P_{\mu}^{X} = 0.531$, $P_{\mu}^{Y} = 0.110$, $P_{\nu}^{X} = 209$, $P_{\nu}^{Y} = 0.043$, $P_{\omega}^{X} = 089$, $P_{\omega}^{Y} = 0.018$.

Then, for a higher practical security, we implement the security analysis in Ref. [56] and three-intensity decoy-state

 Table 3. Experimental Gains and Quantum Bit Error

 Rates of Our MDI-QKD System

$\mu_a \mu_b$	Q^X	E^X	Q^Y	E^{Y}
μμ	1.82×10^{-4}	2.69%	3.40×10^{-4}	26.08%
μυ	4.67×10^{-5}	4.75%	1.19×10^{-4}	36.26%
μω	7.06×10^{-6}	51.02%	1.06×10^{-4}	50.10%
νμ	4.90×10^{-5}	5.02%	1.32×10^{-4}	35.86%
νν	1.11×10^{-5}	3.66%	2.13×10^{-5}	26.16%
$\nu\omega$	2.39×10^{-7}	47.26%	5.63×10^{-6}	50.46%
ωμ	4.04×10^{-6}	50.19%	1.01×10^{-4}	50.40%
ων	9.25×10^{-7}	51.88%	4.80×10^{-6}	50.09%

Table 4. Experimental Gains and Quantum Bit Error

μ_a	Q^X	E^X	Q^{Y}	E^{Y}
μ	3.10×10^{-2}	0.39%	3.09×10^{-2}	0.28%
ν	3.67×10^{-3}	0.38%	3.71×10^{-3}	0.29%
ω	1.96×10^{-4}	2.66%	1.93×10^{-4}	1.82%
μ_b	Q^X	E^X	Q^Y	E^{Y}
μ	3.13×10^{-2}	0.38%	3.14×10^{-2}	0.34%
ν	3.69×10^{-3}	0.50%	3.71×10^{-3}	0.48%
ω	1.95×10^{-4}	1.89%	1.98×10^{-4}	2.09%

method. The Hoeffding's inequality [66] is used for our fluctuation analysis. In our calculation of secure key rate, the failure probability of parameter estimation, ε_{sec} , is equal to 10^{-9} , and the failure probability of error-verification step, ε_{cor} , is equal to 10^{-15} . In addition, the total number of sifted pulse pairs $N_t = 1.16 \times 10^9$.

Finally, we obtain the secure key rates of 6.289×10^{-3} (Alice-Charlie) and 6.155×10^{-3} (Bob-Charlie) for transmission distance of 10 km, as shown in Fig. 3. The gains and QBERs of our BB84 QKD systems are shown in Table 4. The method to calculate the secure key rate is shown in Appendix C.

C. Summary

According to the results shown in the section MDI-QKD and section BB84 QKD, we summarize the performance of our system as network link rates in Fig. 3.

In summary, our system implements a compatibility between two distinct protocols: BB84 and MDI. The finite-sized effect is included in our analysis for the requirement of practical security. Moreover, the three-intensity decoy-state method is used. Based on the FM structure, our BB84 subsystems can automatically compensate for the channel polarization disturbance. The secure key rate is about 250 kbps at 10 km of fiber. Also, our new scheme can realize low-error-rate phase-encoding MDI-QKD without the optical switch and phase-postselection techniques. The system complexities of Alice and Bob are reduced, whose sensitivity to cost is higher than Charlie's. And we obtain 410 bps of secure key rate with a high level of practical security.

Rates of Our BB84 QKD Systems

In conclusion, with the help of our new scheme, the incompatibilities between phase-encoding MDI and BB84 protocols are removed. These protocols are integrated into a single system to introduce the advantages of the NSA network so that our system can switch the engaged protocol between them as required and further bridges the gap between existing phaseencoding BB84 networks and MDI networks.

BB84 is one of the most widely used protocols in trusted node networks, and MDI is an ideal candidate for the untrusted-node-based network. These features certainly show an evolution path toward MDI networks. The NSA-MDI scheme can immediately make the phase-encoding BB84 networks support MDI and is not limited to a specific realization of phase encoding. The dependence of networks on node credibility is also lowered; thus the network survivability is improved. Moreover, various requirements of different application scenarios, especially the requirements of high key rate or high security level can be satisfied in one network. More importantly, all these advantages can be obtained without hardware cost, which benefits from the coordination of our scheme.

The network is the final form of QKD application and the first step toward the quantum Internet. During the popularization and upgrade of QKD, the cost needs to be lowered. Our NSA network scheme provides an evolution path that targets both cost and demand. Existing phase-encoding BB84 networks and production lines are fully exploited, and the lower threshold and higher usability can quicken the construction of MDI-QKD networks.

APPENDIX A: MDI-QKD WITH PHASE-RANDOMIZED COHERENT STATES

In this section, we show in detail the evolution of the quantum state in our system according to the checkpoints marked in Fig. 4.

Checkpoint 1. Alice and Bob prepare coherent states with intensities μ_a and μ_b , respectively, and randomize the phases. The initial joint state is

$$|e^{i\phi_a}\sqrt{\mu_a}\rangle_a|e^{i\phi_b}\sqrt{\mu_b}\rangle_a,$$
 (A1)

where ϕ_a and ϕ_b are the overall randomized phases.

Checkpoint 2. The pulses are split into two orthogonal optical modes (*l* mode and *s* mode),

$$\left|e^{i\phi_{a}}\sqrt{\frac{\mu_{a}}{2}}\right\rangle_{a_{l}}\left|e^{i(\phi_{a}+\theta_{a})}\sqrt{\frac{\mu_{a}}{2}}\right\rangle_{a_{s}}\left|e^{i\phi_{b}}\sqrt{\frac{\mu_{b}}{2}}\right\rangle_{b_{l}}\left|e^{i(\phi_{b}+\theta_{b})}\sqrt{\frac{\mu_{b}}{2}}\right\rangle_{b_{s}},$$
(A2)

where θ_{a} and θ_{b} are the relative phases between the two modes, which are modulated by Alice and Bob, respectively.

Checkpoint 3. After passing through lossy channels, the joint state can be expressed by

$$\left| e^{i\phi_{a}} \sqrt{\frac{\mu_{a}\eta_{a}}{2}} \right\rangle_{a_{l}} \left| e^{i(\phi_{a}+\theta_{a})} \sqrt{\frac{\mu_{a}\eta_{a}}{2}} \right\rangle_{a_{s}} \\ \otimes \left| e^{i\phi_{b}} \sqrt{\frac{\mu_{b}\eta_{b}}{2}} \right\rangle_{b_{l}} \left| e^{i(\phi_{b}+\theta_{b})} \sqrt{\frac{\mu_{b}\eta_{b}}{2}} \right\rangle_{b_{s}},$$
 (A3)





where η_a and η_b are the channel transmittances of Alice-Charlie and Bob-Charlie.

Checkpoint 4. The first BS and PM of Charlie transform the states into

$$\left| e^{i\phi_{a}} \sqrt{\frac{\mu_{a}\eta_{a}}{2}} - e^{i\phi_{b}} \sqrt{\frac{\mu_{b}\eta_{b}}{2}} \right\rangle_{ll}$$

$$\otimes \left| e^{i(\phi_{a}+\theta_{a}+\theta_{c})} \sqrt{\frac{\mu_{a}\eta_{a}}{2}} + e^{i(\phi_{b}+\theta_{b}+\theta_{c})} \sqrt{\frac{\mu_{b}\eta_{b}}{2}} \right\rangle_{ss}$$

$$\otimes \left| e^{i(\phi_{a}+\theta_{c})} \sqrt{\frac{\mu_{a}\eta_{a}}{2}} + e^{i(\phi_{b}+\theta_{c})} \sqrt{\frac{\mu_{b}\eta_{b}}{2}} \right\rangle_{ls}$$

$$\otimes \left| e^{i(\phi_{a}+\theta_{a})} \sqrt{\frac{\mu_{a}\eta_{a}}{2}} - e^{i(\phi_{b}+\theta_{b})} \sqrt{\frac{\mu_{b}\eta_{b}}{2}} \right\rangle_{sl}, \quad (A4)$$

where θ_c is the modulated phase of Charlie.

Checkpoint 5. Before the pulses arrive at SPDs, the states are changed to

$$\left|\frac{\sqrt{\mu_{a}\eta_{a}}}{2\sqrt{2}}\left(e^{i(\phi_{a}+\theta_{c})}+e^{i(\phi_{a}+\theta_{a})}\right)+\frac{\sqrt{\mu_{b}\eta_{b}}}{2\sqrt{2}}\left(e^{i(\phi_{b}+\theta_{c})}-e^{i(\phi_{b}+\theta_{b})}\right)\right\rangle_{D_{1}} \\ \otimes\left|\frac{\sqrt{\mu_{a}\eta_{a}}}{2\sqrt{2}}\left(e^{i(\phi_{a}+\theta_{c})}-e^{i(\phi_{a}+\theta_{a})}\right)+\frac{\sqrt{\mu_{b}\eta_{b}}}{2\sqrt{2}}\left(e^{i(\phi_{b}+\theta_{c})}+e^{i(\phi_{b}+\theta_{b})}\right)\right\rangle_{D_{2}}.$$
(A5)

For simplicity, we use $|\psi_1\rangle_{D_1}$ and $|\psi_2\rangle_{D_2}$ to replace the expression above,

$$|\psi_1\rangle_{D_1}\otimes|\psi_2\rangle_{D_2}.$$
 (A6)

Then, the response probabilities of D_1 and D_2 can be obtained by

$$p_{\mu}^{D_1} = 1 - (1 - Y_0)(1 - P_{ap}) \exp(-|\psi_1|^2),$$

$$p_{\mu}^{D_2} = 1 - (1 - Y_0)(1 - P_{ap}) \exp(-|\psi_2|^2),$$
(A7)

where Y_0 is the dark-count rate and P_{ap} is the after-pulse rate of detectors [67],

$$\begin{split} |\psi_{1}|^{2} &= \frac{\mu_{a}\eta_{a}}{4} (1 + \cos(\theta_{c} - \theta_{a})) + \frac{\mu_{b}\eta_{b}}{4} (1 - \cos(\theta_{c} - \theta_{b})) \\ &+ \frac{\sqrt{\mu_{a}\eta_{a}\mu_{b}\eta_{b}}}{4} (\cos(\phi_{a} - \phi_{b}) + \cos(\phi_{a} - \phi_{b} + \theta_{a} - \theta_{c}) \\ &- \cos(\phi_{b} - \phi_{a} + \theta_{b} - \theta_{c}) - \cos(\phi_{a} - \phi_{b} + \theta_{a} - \theta_{b})) \\ |\psi_{2}|^{2} &= \frac{\mu_{a}\eta_{a}}{4} (1 - \cos(\theta_{c} - \theta_{a})) + \frac{\mu_{b}\eta_{b}}{4} (1 + \cos(\theta_{c} - \theta_{b})) \\ &+ \frac{\sqrt{\mu_{a}\eta_{a}\mu_{b}\eta_{b}}}{4} (\cos(\phi_{a} - \phi_{b}) + \cos(\phi_{a} - \phi_{b} + \theta_{c} - \theta_{b})) \\ &- \cos(\phi_{b} - \phi_{a} + \theta_{c} - \theta_{a}) - \cos(\phi_{a} - \phi_{b} + \theta_{a} - \theta_{b})). \end{split}$$
(A8)

For simplicity, we use the following notations:

$$A = \frac{\sqrt{\mu_a \eta_a}}{2}, B = \frac{\sqrt{\mu_b \eta_b}}{2}.$$
 (A9)

Then, the detection intensities can be simplified and shown in Table 5. Here, without loss of generality, we let $\theta_c = 0$ for simplicity. The physical meaning of θ_c is the reference phase of θ_a and θ_b . Therefore, the phase shifting between θ_a and θ_b can be compensated for by the modulation of θ_c .

Table 5. Detection Intensities on X and Y Basis

θ_a	θ_b	$ \psi ^2$
0	0	$ \psi_1 ^2 = 2A^2$
π	π	$ \psi_2 ^2 = 2B^2$ $ \psi_1 ^2 = 2A^2$
0	π	$ \psi_2 ^2 = 2B^2$ $ \psi_1 ^2 = 2A^2 + 2B^2 + 4AB\cos(\phi_a - \phi_b)$
π	0	$ \psi_2 ^2 = 0$ $ \psi_1 ^2 = 0$
$\frac{\pi}{2}$	$\frac{\pi}{2}$	$ \psi_2 ^2 = 2A^2 + 2B^2 + 4AB\cos(\phi_a - \phi_b)$ $ \psi_1 ^2 = A^2 + B^2 - 2AB\sin(\phi_a - \phi_b)$
$\frac{3\pi}{2}$	$\frac{3\pi}{2}$	$ \psi_2 ^2 = A^2 + B^2 + 2AB \sin(\phi_a - \phi_b) \psi_1 ^2 = A^2 + B^2 + 2AB \sin(\phi_a - \phi_b)$
<u>π</u>	<u>3π</u>	$ \psi_2 ^2 = A^2 + B^2 - 2AB\sin(\phi_a - \phi_b)$ $ \psi_1 ^2 = A^2 + B^2 + 2AB\cos(\phi_a - \phi_b)$
2	2	$ \psi_2 ^2 = A^2 + B^2 + 2AB \cos(\phi_a - \phi_b)$
<u>3π</u> 2	$\frac{\pi}{2}$	$\begin{aligned} \psi_1 ^2 &= A^2 + B^2 + 2AB \cos(\phi_a - \phi_b) \\ \psi_2 ^2 &= A^2 + B^2 + 2AB \cos(\phi_a - \phi_b) \end{aligned}$

The valid response is defined as the coincidence of the clicks of D_1 and D_2 . Therefore, the response probabilities of the X basis can be given by

$$\begin{aligned} Q^X_{\mu}|_{\theta_a=\theta_b} &= \sum_{\theta_a=\theta_b\in\{0,\pi\}} \frac{1}{4\pi^2} \int_0^{2\pi} \int_0^{2\pi} p^{D_1}_{\mu} p^{D_2}_{\mu} d\phi_a d\phi_b \\ &= 2(1-(1-Y_0)(1-P_{\rm ap})(2-2A^2-2B^2)) \\ &+ (1-Y_0)^2(1-P_{\rm ap})^2(1-2A^2)(1-2B^2)), \end{aligned}$$
$$\begin{aligned} Q^X_{\mu}|_{\theta_a\neq\theta_b} &= \sum_{\theta_a\neq\theta_b\in\{0,\pi\}} \frac{1}{4\pi^2} \int_0^{2\pi} \int_0^{2\pi} p^{D_1}_{\mu} p^{D_2}_{\mu} d\phi_a d\phi_b \\ &= 2(1-(1-Y_0)(1-P_{\rm ap})(2-2A^2-2B^2)) \\ &+ (1-Y_0)^2(1-P_{\rm ap})^2(1-2A^2-2B^2)). \end{aligned}$$
 (A10)

Finally, the gains and QBER are given by

$$Q_{\mu}^{X} = Q_{\mu}^{X}|_{\theta_{a}=\theta_{b}} + Q_{\mu}^{X}|_{\theta_{a}\neq\theta_{b}},$$

$$E_{\mu}^{X}Q_{\mu}^{X} = e_{d}Q_{\mu}^{X}|_{\theta_{a}=\theta_{b}} + (1 - e_{d})Q_{\mu}^{X}|_{\theta_{a}\neq\theta_{b}},$$
 (A11)

where e_d is the misalignment-error rate.

Similarly, the gains and QBER of the Y basis can be given by

$$\begin{split} Q^{Y}_{\mu}|_{\theta_{a}=\theta_{b}} &= \sum_{\substack{\theta_{a}=\theta_{b}\in\{\frac{\pi}{2},\frac{3\pi}{2}\}}} \frac{1}{4\pi^{2}} \int_{0}^{2\pi} \int_{0}^{2\pi} p_{\mu}^{D_{1}} p_{\mu}^{D_{2}} \mathrm{d}\phi_{a} \mathrm{d}\phi_{b} \\ &= 2(1-(1-Y_{0})(1-P_{\mathrm{ap}})(2-2A^{2}-2B^{2}) \\ &+ (1-Y_{0})^{2}(1-P_{\mathrm{ap}})^{2}((1-A^{2}-B^{2})^{2}-2A^{2}B^{2})), \\ Q^{Y}_{\mu}|_{\theta_{a}\neq\theta_{b}} &= \sum_{\substack{\theta_{a}\neq\theta_{b}\in\{\frac{\pi}{2},\frac{3\pi}{2}\}}} \frac{1}{4\pi^{2}} \int_{0}^{2\pi} \int_{0}^{2\pi} p_{\mu}^{D_{1}} p_{\mu}^{D_{2}} \mathrm{d}\phi_{a} \mathrm{d}\phi_{b} \end{split}$$

$$= 2(1 - (1 - Y_0)(1 - P_{ap})(2 - 2A^2 - 2B^2) + (1 - Y_0)^2(1 - P_{ap})^2((1 - A^2 - B^2)^2 + 2A^2B^2)),$$
(A12)

$$Q_{\mu}^{Y} = Q_{\mu}^{Y}|_{\theta_{a}=\theta_{b}} + Q_{\mu}^{Y}|_{\theta_{a}\neq\theta_{b}},$$

$$E_{\mu}^{Y}Q_{\mu}^{Y} = e_{d}Q_{\mu}^{Y}|_{\theta_{a}=\theta_{b}} + (1 - e_{d})Q_{\mu}^{Y}|_{\theta_{a}\neq\theta_{b}}.$$
 (A13)

APPENDIX B: CALCULATION OF SECURE KEY RATE FOR MDI-QKD

In this section, we use the method of Ref. [57] to calculate the secure key rate, which treats the statistical fluctuation with Chernoff's bounds. It is enough to show the feasibility of our scheme, although it is not a complete finite-sized analysis against the coherent attack. The secure key rate R can be obtained by

$$R = P_{\mu}^{X^{2}}(\mu^{2}e^{-2\mu}Y_{11}^{X,L}(1 - H_{2}(e_{11}^{Y,U})) - Q_{\mu_{a}\mu_{b}}^{X}f_{e}H_{2}(E_{\mu_{a}\mu_{b}}^{X}),$$
(B1)

where P_{μ}^{X} is the probability that Alice and Bob send μ state with the X basis, $Y_{11}^{X,L}$ is the lower bound of the yield of singlephoton pairs, $e_{11}^{X,U}$ is the upper bound of the phase-flip error rate, $Q_{\mu\mu}^{X}$ and $E_{\mu\mu}^{X}$ are the observed gain and QBER that both Alice and Bob send μ state with the X basis, $H_2(x) = -x\log_2(x) - (1-x)\log_2(1-x)$ is the binary Shannon entropy function, and f_e is the error correction efficiency.

The yield and phase-flip error rate can be estimated by observables according to [57]

$$Y_{11}^{X,L} = \frac{1}{(\mu_a - \omega_a)(\mu_b - \omega_b)(\nu_a - \omega_a)(\mu_b - \omega_b)(\mu_a - \omega_a)} \times ((\mu_a^2 - \omega_a^2)(\mu_b - \omega_b)(Q_{\nu_a\nu_b}^{X,L}e^{(\nu_a + \nu_b)} + Q_{\omega_a\omega_b}^{X,L}e^{(\omega_a + \omega_b)} - Q_{\nu_a\omega_b}^{X,U}e^{(\omega_a + \nu_b)}) - (\nu_a^2 - \omega_a^2)(\nu_b - \omega_b)(Q_{\mu_a\mu_b}^{X,U}e^{(\mu_a + \mu_b)} + Q_{\omega_a\omega_b}^{X,U}e^{(\omega_a + \omega_b)} - Q_{\omega_a\mu_b}^{X,L}e^{(\omega_a + \mu_b)})),$$

$$- Q_{\mu_a\omega_b}^{X,L}e^{(\mu_a + \omega_b)} - Q_{\omega_a\mu_b}^{X,L}e^{(\omega_a + \mu_b)})),$$
(B2)

$$e_{11}^{Y,U} = \frac{1}{(\nu_a - \omega_a)(\nu_b - \omega_b)Y_{11}^{Y,L}} \times (e^{(\nu_a + \nu_b)}EQ_{\nu_a\nu_b}^{Y,U} + e^{(\omega_a\omega_b)}EQ_{\omega_a\omega_b}^{Y,U} - e^{(\nu_a + \omega_b)}EQ_{\nu_a\omega_b}^{Y,L} - e^{(\omega_a + \nu_b)}EQ_{\omega_a\omega_b}^{Y,L}),$$
(B3)

$$Y_{11}^{Y,L} = \frac{1}{(\mu_a - \omega_a)(\mu_b - \omega_b)(\nu_a - \omega_a)(\mu_b - \omega_b)(\mu_a - \omega_a)} \times ((\mu_a^2 - \omega_a^2)(\mu_b - \omega_b)(Q_{\nu_a\nu_b}^{Y,L}e^{(\nu_a + \nu_b)} + Q_{\omega_a\omega_b}^{Y,L}e^{(\omega_a + \omega_b)} - Q_{\nu_a\omega_b}^{Y,U}e^{(\nu_a + \omega_b)} - Q_{\omega_a\nu_b}^{Y,U}e^{(\omega_a + \nu_b)}) - (\nu_a^2 - \omega_a^2)(\nu_b - \omega_b)(Q_{\mu_a\mu_b}^{Y,U}e^{(\mu_a + \mu_b)} + Q_{\omega_a\omega_b}^{Y,U}e^{(\omega_a + \omega_b)} - Q_{\mu_a\omega_b}^{Y,L}e^{(\omega_a + \mu_b)})),$$
(B4)

where $Q_{\alpha_a \alpha_b}^{\beta,\chi}$ and $E Q_{\alpha_a \alpha_b}^{\beta,\chi}$ are the χ bounds of the observable gain and error rate that Alice sends α_a state and Bob sends α_b state with β basis, respectively, $\beta \in \{X, Y\}, \chi \in \{U, L\}, \alpha \in \{\mu, \nu, \omega\}.$

In order to deal with the statistical fluctuation, the observables can be bounded by Chernoff's bounds,

Table 6. Parameters of Secure Key Calculation in Our MDI-QKD System

f_{e}	Y_0	η_d	ε	μ	ν	ω
1.16	7.5×10^{-6}	25%	10^{-10}	0.284	0.057	0

$$Q_{\alpha_{a}\alpha_{b}}^{\beta,U} = Q_{\alpha_{a}\alpha_{b}}^{\beta} \left(1 + \frac{f((\varepsilon/2)^{4}/16)}{\sqrt{N_{\alpha_{a}\alpha_{b}}^{\beta}Q_{\alpha_{a}\alpha_{b}}^{\beta}}}\right)$$
$$Q_{\alpha_{a}\alpha_{b}}^{\beta,L} = Q_{\alpha_{a}\alpha_{b}}^{\beta} \left(1 - \frac{f((\varepsilon/2)^{3/2})}{\sqrt{N_{\alpha_{a}\alpha_{b}}^{\beta}Q_{\alpha_{a}\alpha_{b}}^{\beta}}}\right)$$
$$EQ_{\alpha_{a}\alpha_{b}}^{\beta,U} = EQ_{\alpha_{a}\alpha_{b}}^{\beta} \left(1 + \frac{f((\varepsilon/2)^{4}/16)}{\sqrt{N_{\alpha_{a}\alpha_{b}}^{\beta}EQ_{\alpha_{a}\alpha_{b}}^{\beta}}}\right)$$
$$EQ_{\alpha_{a}\alpha_{b}}^{\beta,L} = EQ_{\alpha_{a}\alpha_{b}}^{\beta} \left(1 - \frac{f((\varepsilon/2)^{3/2})}{\sqrt{N_{\alpha_{a}\alpha_{b}}^{\beta}EQ_{\alpha_{a}\alpha_{b}}^{\beta}}}\right),$$
(B5)

where ε is the failure probability of statistical fluctuation.

The parameters used to calculate the secure key are shown in Table 6.

APPENDIX C: CALCULATION OF SECURE KEY RATE FOR BB84 QKD

The secure key rate R can be generated by

$$R = \frac{1}{N} \left(s_0^X + s_1^X (1 - H_2(e_{1,p}^X)) - \lambda_{\rm EC} - 6\log_2 \frac{21}{\varepsilon_{\rm sec}} - \log_2 \frac{2}{\varepsilon_{\rm cor}} \right),$$
(C1)

where s_0 is the number of vacuum events, s_1 is the number of single-photon events, $e_{1,p}$ is the phase error rate, N is the total number of pulses (sent by Alice or Bob), $\beta \in \{X, Y\}$ represents a basis, $H_2(x) = -x \log_2(x) - (1 - x) \log_2(1 - x)$ is the binary Shannon entropy function, $\lambda_{\text{EC}} = n^{\beta} f_e H_2(E^{\beta})$ is the consumption of the information in error correction, f_e is the efficiency factor of the error-correction method used, and ε_{cor} and ε_{sec} are secure parameters.

All needed parameters can be estimated by analytic formulas [56,61]. Specifically, the analytic formulas of three-intensity (μ, ν_1, ν_2) decoy scheme are given by

$$s_0^{\omega} = \frac{\tau_0}{\nu_1 - \nu_2} \left(\frac{e^{\nu_2} \nu_1 \eta_{\nu_2}^{\beta, U}}{P_{\nu_2}} - \frac{e^{\nu_1} \nu_2 \eta_{\nu_1}^{\beta, L}}{P_{\nu_1}} \right),$$
(C2)

$$s_{1}^{\omega} = \frac{\mu \tau_{1}}{\mu \nu_{1} - \mu \nu_{2} - \nu_{1}^{2} + \nu_{2}^{2}} \left(\frac{e^{\nu_{1}} n_{\nu_{1}}^{\beta,L}}{P_{\nu_{1}}} - \frac{e^{\nu_{2}} n_{\nu_{2}}^{\beta,U}}{P_{\nu_{2}}} - \frac{\nu_{1}^{2} - \nu_{2}^{2}}{\mu^{2}} \left(\frac{e^{\mu} n_{\mu}^{\omega,U}}{P_{\mu}} - \frac{s_{0}^{\beta}}{\tau_{0}} \right) \right),$$
(C3)

$$e_{1,p}^{\beta} = \frac{v_1^{\tilde{\beta}}}{s_1^{\tilde{\beta}}} + \gamma \left(\varepsilon_{\text{sec}}, \frac{v_1^{\tilde{\beta}}}{s_1^{\tilde{\beta}}}, s_1^{\tilde{\beta}}, s_1^{\beta} \right),$$
(C4)

where

Table 7. Parameters of Secure Key Calculation in OurBB84 QKD System

f _e	Y ₀	η_d	$\boldsymbol{\varepsilon}_{\mathrm{sec}}$	$\boldsymbol{\varepsilon}_{\mathrm{cor}}$	μ	ν	ω
1.16	7.5×10^{-6}	25%	10^{-9}	10 ⁻¹⁵	0.538	0.063	0.003

$$\gamma(a,b,c,d) = \sqrt{\frac{(c+d)(1-b)b\ln 2}{cd}} \sqrt{\log_2\left(\frac{c+d}{cd(1-b)b}\frac{21^2}{a^2}\right)},$$
(C5)

$$\nu_1^{\beta} = \frac{\tau_1}{\nu_1 - \nu_2} \left(\frac{e^{\nu_1} m_{\nu_1}^{\beta, U}}{P_{\nu_1}} - \frac{e^{\nu_2} m_{\nu_2}^{\beta, L}}{P_{\nu_2}} \right).$$
(C6)

 β and $\bar{\beta}$ are different bases, i.e., $\beta = X$ when $\bar{\beta} = Y$ and vice versa. $n_{\alpha}^{\beta,\chi}$ and $m_{\alpha}^{\beta,\chi}$ are the χ bounds of the number of detections and bit error of basis β and intensity α , respectively, $\chi \in \{U, L\}, \alpha \in \{\mu, \nu, \omega\}.$

In order to deal with the statistical fluctuation, according to the counterfactual protocol proposed in Ref. [56], the counts and errors can be bounded by Hoeffding's inequality,

$$n_{\alpha}^{\beta,U} = n_{\alpha}^{\beta} + \sqrt{\frac{n^{\beta}}{2} \ln \frac{21}{\varepsilon_{sec}}},$$

$$n_{\alpha}^{\beta,L} = n_{\alpha}^{\beta} - \sqrt{\frac{n^{\beta}}{2} \ln \frac{21}{\varepsilon_{sec}}},$$

$$m_{\alpha}^{\beta,U} = m_{\alpha}^{\beta} + \sqrt{\frac{m^{\beta}}{2} \ln \frac{21}{\varepsilon_{sec}}},$$

$$m_{\alpha}^{\beta,L} = m_{\alpha}^{\beta} - \sqrt{\frac{m^{\beta}}{2} \ln \frac{21}{\varepsilon_{sec}}},$$
(C7)

where n_{α}^{ω} and m_{α}^{ω} are the number of detections and bit error of basis ω and intensity α observed in experiment.

The parameters used in our secure key calculation are shown in Table 7.

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