

# Inverter power supply and control system for high pulse energy laser shock processing

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In order to meet high pulse laser output, the laser shock power supply based on insulated gate bipolar transistor (IGBT) inverter technology, constant current-constant voltage charger mode, high voltage pulse spark, and high pulse current discharger technology is designed. The master oscillator stage and two amplifier stages for xenon flash lamp spark and discharge circuits are designed in this power supply. The voltage of energy storage capacitors can be adjusted from 1000 to 3000 V. A variety of measures, such as FREE and  $Q$ -switch mode, trigger signal delay, water confining layer control, cooling water control are provided for laser shock processing technology optimization.

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Laser shock processing (LSP) technology has been widely introduced in foreign aviation industry and other industrial fields<sup>[1-4]</sup>. With wide application prospects, LSP technology has been rapidly developed in domestic aviation industry.

In order to obtain intensive pulse energy laser of tens joules or even hundreds joules, Nd: glass laser was usually adopted<sup>[1]</sup>. Although the operating frequency of Nd: glass laser is relatively low, it still has the obvious advantage in the field of LSP for the high pulse laser energy.

The key technologies of Nd: glass laser for LSP include stable charging technology, xenon flash lamp high voltage trigger technology, large current discharge technology,  $Q$ -switch and xenon flash lamp trigger synchronization control technology<sup>[5]</sup>.

Commonly, 50-Hz alternating current (AC) was adjusted by thyristors in early domestic Nd: glass laser power supply, then, the AC was promoted to high voltage by step-up transformer. At the beginning of charge, the large impact current was easy to result in the damage of charging circuit components. Besides the charge noise, the charging efficiency was very low. The charge process was finished at fully charge. For leakage current of energy storage capacitors, output pulse laser energy was different at different moment of discharge trigger. This charge mode made the output pulse laser energy fluctuate.

For the xenon flash lamp trigger circuit was susceptible to electromagnetism interference, the unexpected trigger was always arisen. In addition, actual LSP technique requirements have not been considered in the early domestic laser shock equipment.

Recently, with the emerging of new power parts and control technologies, domestic inverter technology has been rapidly developed. It is possible to manufacture a new type of laser shock equipment with high pulse energy.

With Nd: glass laser introduced from abroad, a new type laser shock power supply with high pulse energy and its control system, which based on the insulated gate bipolar transistor (IGBT) inverter technology, constant current-constant voltage charge technology and high voltage pulse triggering technology, have been developed. Discharge pulse with 500- $\mu$ s pulse width and 5000-A peak current is obtained. FREE mode and  $Q$ -switch mode are provided in control system. It also provides a series of optimization LSP measures in this control system, such as automatic control of water confining layer and cooling water. By these measures adopted, the automatization level of laser shock processing system has been improved.

The charge and discharge circuits are included in main circuit topology. The number of charge and discharge circuits is related with the number of flash lamps in laser optical system. The Nd: glass laser optical path is showed in Fig. 1. It includes the master oscillator stage and the amplifier stages.

Three Nd: glass crystal rods can be seen from Fig. 1. It means that three charge and discharge circuits are needed to drive three flash lamps respectively. The charge and discharge circuits adopted to spark flash lamp of the Nd: glass rod 1 are known as the master oscillator stage. Those circuits adopted to spark flash lamps for Nd: glass rod 2 and rod 3 are called amplifier stage.

The constant current-constant voltage inverter charge circuit of master oscillator stage is shown in Fig. 2. Those charge circuits of amplifier stage are similar to the circuit showed in Fig. 2.

The rectifier circuit MD<sub>1</sub>, inductance  $L_{01}$ , half-bridge inverter circuit (including capacitor  $C_{01}$ ,  $C_{02}$  and IGBT  $Q_{01}$ ,  $Q_{02}$ ), high frequency step-up transformer B<sub>0</sub>, secondary rectifier circuit (including  $D_{01}$ ,  $D_{02}$  high voltage rectifier diodes) had been introduced in master oscillator stage's constant current-constant voltage circuit. The

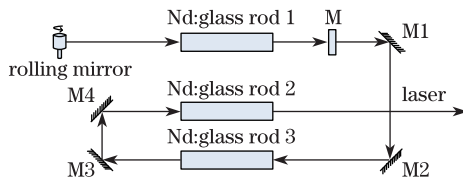


Fig. 1. Optical paths of Nd: glass laser.

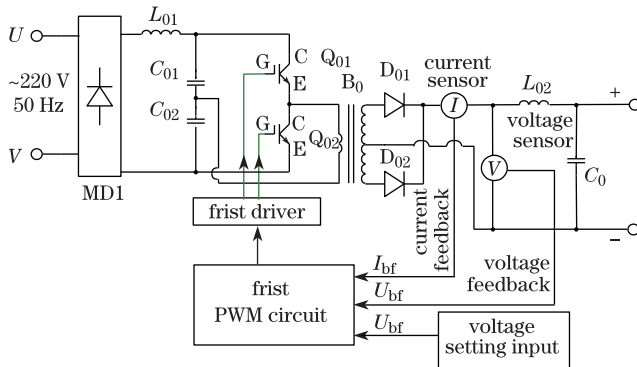


Fig. 2. Constant current-constant voltage circuit topology.

capacity of the storage capacitor  $C_0$  is  $700 \mu\text{F}$ , which is formed by series-parallel  $150 \mu\text{F}/1100 \text{V}$  capacitors.

The circuit operation process is as follows. Through the rectifier filter circuit MD1 and  $L_{01}$ ,  $220 \text{V}/50 \text{Hz}$  AC is changed to  $310\text{-V}$  stable direct current (DC). This DC is transformed to  $20\text{-kHz}$  square AC by half-bridge inverter circuit (includes capacitor  $C_{01}$ ,  $C_{02}$ , and IGBT  $Q_{01}$ ,  $Q_{02}$ ). The  $20\text{-kHz}$  square wave AC coupled by step-up transformer  $B_0$  is transformed to high voltage AC with same frequency. This high voltage AC is the input of the secondary rectifier circuit and inductor  $L_{02}$ . Through these rectifier and filter parts, DC can be achieved to the need of the storage capacitor  $C_0$  charging. The charge voltage can be continuously adjusted from  $0$  to  $3000 \text{V}$ . According to the difference between the detection voltage of storage capacitor  $C_0$  and enactment voltage value, the constant current or constant voltage charge mode will be chosen by control system.

The circuit of flash lamp trigger and discharge in master oscillator stage is shown in Fig. 3. Those amplifier stage's flash lamp trigger and discharge circuits are similar to the circuit shown in Fig. 3.

In order to protect the discharge circuit components, a  $50\text{-}\mu\text{H}$  hollow inductance is introduced to restrict increase rate of the discharge current. The  $B_1$  is a high-voltage pulse trigger transformer, which can provide tens of thousands volts voltage negative pulse to trigger the flash lamp. The ratio of primary coils and secondary coils is  $1:20$  in the transformer  $B_1$ . When the  $1500\text{-V}$  pulse is inputted in primary coils,  $30000\text{-V}$  high voltage pulse will be inspired in the secondary coils.

The operation process of the discharge circuit is as follows. Firstly, the  $C_1$  is charged to  $1500 \text{V}$ , with  $R_1$  adopted to restrict current peak. When the synchronous trigger signal is valid, switch  $Q_0$  is turned on by drive circuit. The charge stored in  $C_1$  will be discharged by the primary coils of transformer  $B_1$  and switch  $Q_0$ . Thus  $1500\text{-V}$  voltage pulse will be produced at the  $B_1$  primary coils. The  $30000\text{-V}$  high voltage pulse will arise at the  $B_1$  secondary coils for the transformer  $B_1$  coupling. It

makes xenon in the flash lamp to be ionized. Thus, a narrow conductive channel is attained. The charge stored in  $C_0$  will be discharged by the inductor  $L$ , flash lamp, and  $B_1$  secondary coils. With a large number of xenon ionization, electron impact acceleration and xenon rapid heating, conductive channel is increased. After several  $\mu\text{s}$ , the entire section of flash lamp is full of arc. A high pulse discharge current is arisen. This discharge process can be considered as a relative steady state discharge, in which the xenon flash lamp can be considered as very small resistance<sup>[6]</sup>.

In order to restrict the impact current at the initial moment of charge and to reduce circuit losses, constant current-constant voltage charge mode was introduced in the control system. In this charge mode, Charge process is divided into constant current charge stage and constant voltage charge stage. At the stage of constant current charge, the energy storage capacitor is charged fast by a constant current, which makes the capacitor voltage increase linearly. In order to avoid overage charge, the constant current charge stage will be stopped until there is a little difference between the capacitor voltage  $U_1$  and initialization voltage  $U_0$ . The control system will make the charge process turn to constant voltage charge stage. At this charge stage, the charge current is extremely small. When the capacitor voltage is charged to the set value  $U_0$ , the charge current of mA-class is ordered to maintain constant voltage on the energy storage capacitor. Once the discharge trigger signal is received, the constant voltage charge stage will stop.

In order to achieve the constant current and constant voltage output characteristics of the inverter power supply, inner loop of current negative feedback and outer loop of voltage negative feedback controls had been introduced in the control circuit. The inverter power supply output characteristics control sketch map is shown in Fig. 4.

During the storage capacitor  $C_0$  being charged by inverter power supply, the control circuit of current negative feedback closed-loop will be operated at the time of  $C_0$  voltage below enactment voltage. Namely, at the stage of constant current charge, charge current  $I_f$  as feedback parameter participates in the system proportional-integral-derivative (PID) regulator. When the voltage of the storage capacitor  $C_0$  has been charged to enactment voltage, the control system of inverter power supply will be turned to constant voltage charge mode. In this charge stage, capacitor voltage  $U_f$  as the feedback parameter is adjusted by PID regulator, which will make the voltage of  $C_0$  preserve enactment voltage.

After the energy storage capacitor has been charged fully, the discharge process starts immediately at the time of discharge trigger signal appearance. This discharge mode is named FREE mode. Generally, this discharge mode is applied to debug the optical paths. The energy is difficult to reach the threshold for laser output at free discharge state. Usually,  $Q$ -switch technology is adopted to achieve laser output. It means that some ways are used to make  $Q$  value in laser cavity change according to certain time procedures. At the time of pumping startup, the resonance can not be produced for the low  $Q$  at cavity or the high resonance threshold. When the  $Q$  value is suddenly increased, the laser oscillation is set

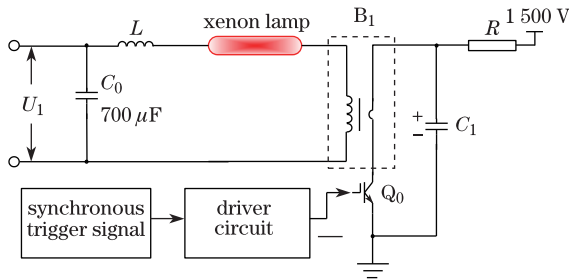


Fig. 3. Flash lamp trigger and discharge circuit.

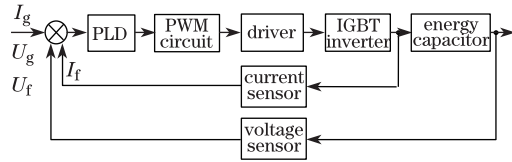


Fig. 4. Control sketch map of charge.

up quickly. In a very short time, a high pulse energy laser is produced at the end of resonance cavity<sup>[7,8]</sup>.

*Q*-switch operation and FREE operation modes had been designed for high pulse energy laser output and the optical path debugging respectively. For the high speed rotating prism tuning *Q* technology is adopted, the synchronous discharge trigger pulse should be considered at control system.

Effective time of ‘delay time’ is named as pulse duty between a active trigger switch signal to the discharge signal. The core integrated circuit (IC) producing the trigger ‘delay time’ is NE555. A single NE555 circuit is shown in Fig. 5(a). A series of NE555 and logic control circuit have been introduced in the trigger synchronization control circuit, which circuit diagram is shown in Fig. 5(b).

In Fig. 5(a), when the pulse falling edge by  $C_3$  is inputted into the NE555 trigger port, a width-adjustable pulse signal will arise at output port. The output pulse width depends on resistor  $R_7$ ,  $VR_1$  and capacitor  $C_6$ . The duration of the output pulse signal can be adjusted by resistance  $VR_1$ . Thus, ‘delay time’ can be adjusted from 0 to 1000  $\mu s$ .

The 2.5-ms discharge button signal duration, 400- $\mu s$  rotating mirror pulse duration, 0-1000- $\mu s$  “delay time”, and 20- $\mu s$  discharger duration signal are designed in this system. The signals of *Q*-switch control, FREE control and cooling water control are 12-V positive signal.

The *Q*-switch signal and FREE signal are complementary, which means one signal is “1” state, and the other signal is “0” state.

At the *Q*-switch state, *Q*-switch signal at “1”, free control signal at “0”, the output signal of U1B is positive. When rotating prism signal and discharger button signal are “1”, the output of U1A is low. A pulse falling edge is inputted into the NE555 tuning circuit 1, then a delay time pulse, which positive pulse width is determined by delay time of NE555, will arise at the NE555 tuning circuit1 output port. At the end of the delay time, namely the falling edge of the delay time signal is inputted into the NE555 tuning circuit 2. Subsequently, a fixed pulse width positive signal will be produced. When the cooling water control signal is “1”, the discharger trigger signal, which is controlled logically by U2A and U2B, will be

produced. FREE mode and *Q*-switch mode trigger sequences had been detected by TPS2024 oscilloscope. These logic signal diagrams are shown in Fig. 6 respectively.

When there is a large pulse current flow, xenon flash lamp without cooling water will be destroyed easily<sup>[9]</sup>. In order to protect the xenon flash lamp, the protection circuit in the discharge trigger control system, which does not make xenon flash lamp spark without cooling water, had been designed. If the cooling water signal is “0”, the discharger signal will always be maintained low signal “0”. This measure guarantees flash lamp will not be sparked by high voltage pulse without cooling water.

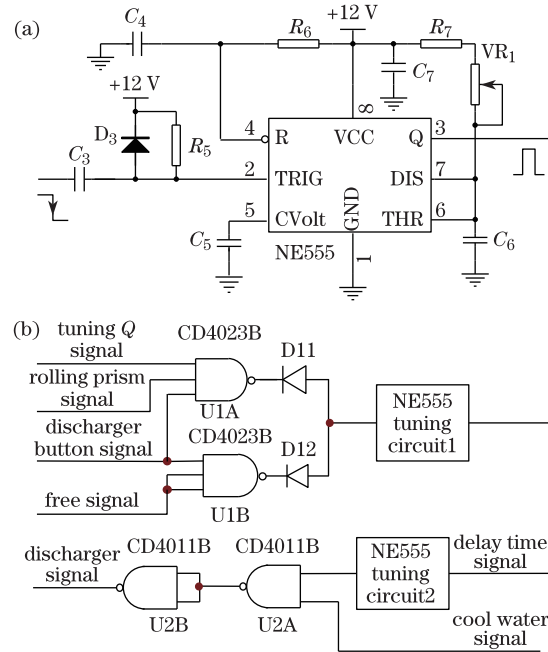
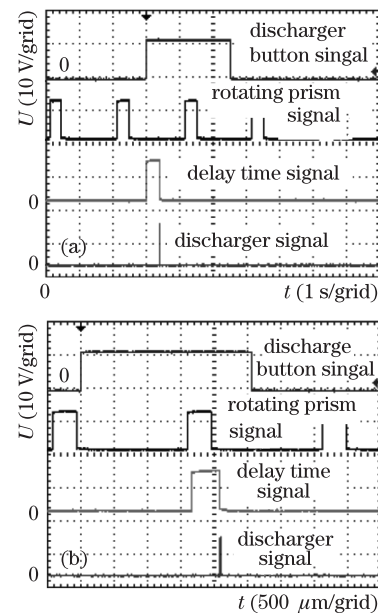


Fig. 5. Trigger logic control circuit. (a) NE555 and its external circuit; (b) trigger timing logic control circuit diagram.

Fig. 6. Discharge trigger sequence. (a) Free mode trigger sequence; (b) *Q*-switch mode trigger sequence.

In addition, the circuit of water confining layer control based on the NE555 was designed.

At the 2900-V  $U_g$ , the voltage of  $C_0$  was detected. The test results is shown in Fig. 7.

It can be seen from Fig. 7 that charge time of the voltage increased from 0 to 2900 V is only 7.6 s. This charge rate can meet the need of the 0.1 Hz discharge frequency.

At the same time, the PWM<sub>1</sub> and PWM<sub>2</sub> pulse signals driving  $Q_{01}$  and  $Q_{02}$  were detected during the constant current charge stage. These signals are shown in Fig. 8(a). At the constant voltage charge stage, PWM<sub>1</sub> and PWM<sub>2</sub> for driving  $Q_{01}$  and  $Q_{02}$  are shown as in Fig. 8(b).

With 1:1000 current transformer and 82- $\Omega$  current sample resistance, the discharge currents, which storage capacitor charge voltage amplitude was set from 2000 to 3000 V, were detected. According to these test results, it can be found that discharge current is increased linearly with the charge voltage of energy storage capacitor increasing. At 3000-V charge voltage, the discharge current wave form is shown in Fig. 9.

It can be seen that discharge current pulse width is 500  $\mu$ s in the Fig. 9. The maximal discharge current amplitude is up to 5000 A.

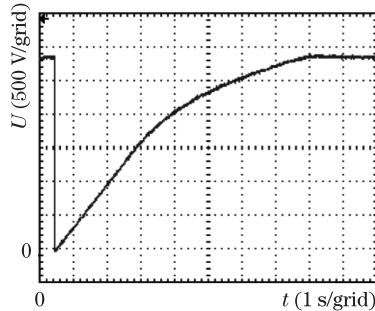


Fig. 7. Charge voltage curve of energy storage capacitor.

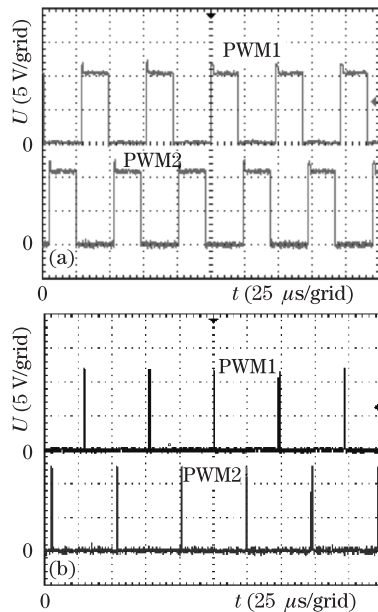


Fig. 8. PWM constant current-constant voltage charge. (a) PWM pulse of constant current charge phase; (b) PWM pulse of constant voltage charge phase.

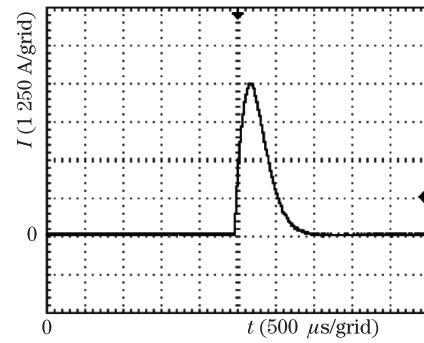


Fig. 9. Discharge current.

**Table 1. Relationship Between Amplifier Stage Voltage and Pulse Laser Energy**

Master Oscillator Stage	Amplifier Stage	Laser Energy
$U_1$ (V)	$U_2$ (V)	$E$ (J)
3007	2007	19.97
3011	2112	23.08
3002	2215	26.85
3005	2310	29.13
3012	2408	32.09
3009	2500	35.23
3003	2609	37.07
3013	2710	40.83
3008	2812	44.05
3010	2909	47.15
3002	3011	49.86

The 3% spectroscope and laser energy test apparatus were used to detect the laser output energy at different charge voltage amplitude. Because the low charge voltage at the master oscillator stage resulted in low pulse laser energy output, the charge voltage of master oscillator stage was set at 3000 V. The charge voltage of the amplifier stage was increased from 2000 to 3000 V, with 100-V increments every times. The different charge voltage amplitudes of amplification stage and its corresponding to laser energy are shown in Table 1. It can be seen from Table 1 that laser energy is enhanced linearly as amplifier stage voltage increases. The laser energy is 49.86 J at 3000-V charge voltage of amplifier stage.

When the master oscillator stage charge voltage and the amplifier stage charge voltage were set at 3000 and 2900 V respectively, the pulse laser energy had been detected 8 times consecutively. The average output laser energy was 47.02 J. The maximum error was 0.5 J between the maximum output energy 47.15 J and the minimum output energy 46.65 J. It is indicated that the power supply performance is relatively stable, which ensures the residual stress on the workpiece surface is very uniform during the continuous operating condition. Thus, the stability of the laser shock quality is enhanced.

In conclusion, constant current-constant voltage charge control strategy and IGBT inverter technology are adopted in this system, which can improve LSP power supply charge speed and stability. The high voltage pulse trigger xenon flash lamp technology is adopted, which

not only ensures reliable trigger to the xenon flash lamp, but also helps to improve the life of the xenon flash lamp. Besides, control and protection circuit based on the NE555 can improve reliability and flexibility of control system, it can enhance the automatization level of the LSP system, which helps to the production efficiency.

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