Design of Optimum Beam Divergence Angle for Intersatellite Optical Communication Systems with Pointing Errors

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Abstract Pointing errors can reduce the number of received signals. The influence of pointing errors on intersatellite optical communication systems is decreased by selecting an optimum laser beam divergence angle for optimizing link reliability in the presence of pointing errors. Given that bias errors are part of pointing errors, an analytic expression of link reliability is obtained in the absence of bias errors. By maximizing link reliability, an analytic expression of the optimum beam divergence angle can be obtained. Bias errors are considered in maximizing link reliability. The numerical results and analytic approximation of the optimum ratio of beam divergence angle to pointing errors for a desired bit rate are obtained. With the application of the designed optimum divergence angle for intersatellite optical communication systems, good system performance can be achieved even in the presence of pointing errors.

Key words free space optical communication; optimum beam divergence angle; link reliability optimization; pointing error

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星间光通信中跟瞄误差下最优光束发散角设计

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摘要 跟瞄误差会引起接收光信号的衰减。为了减小跟瞄误差对星间光通信系统的影响,在跟瞄误差存在的条件下,通过选取最优光束发散角来优化链路可靠性。偏置误差是跟瞄误差的一部分,在没有偏置误差存在的情况下获得了链路可靠性的解析表达式,通过将链路可靠性进行最大化设计,获得了最优光束发散角的解析表达式。将偏置误差也考虑进去再对链路可靠性进行最大化设计,对于给定的通信速率获得了光束发散角与跟瞄误差的最佳比值的数值解及其近似解析表达式。在星间光通信中采用所设计的最优光束发散角,可在存在跟瞄误差的情况下获得良好的系统性能。

关键词 自由空间光通信;最优光束发散角;链路可靠性优化;跟瞄误差
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1 Introduction

Laser satellite communication has become especially attractive in recent years^[1-2] and is perceived to be an important element of future satellite communication infrastructure. In maintaining laser communication links, the pointing acquisition and tracking (PAT) system is essential^[3-6]. However,

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factors such as vibration of satellite platforms, relative motion of satellites, electrical noise, and mechanical noise in the tracking loop limit the pointing accuracy of PAT systems. Thus, optical terminals must operate in the presence of pointing errors, which can cause optical intensity fluctuations at receivers and deteriorate link performance. A communication system may experience data loss when instantaneous mispoint losses exceed the value assigned in the link budget. In such cases, link parameter optimization becomes essential in mitigating the effects of pointing errors.

In many existing studies, performance parameters are optimized through optimum beam divergence angles or beam width selection in the presence of pointing errors. In Ref.[7], link reliability, rate, and range tradeoff are optimized through beam width selection in the presence of atmospheric turbulence and misalignment fading. In Ref.[8], an optimum beam width that minimizes outage performance is obtained in the presence of strong turbulence and misalignment fading channels. Meanwhile, in Ref.[9], an optimization procedure is established by finding the optimum beam width that yields the minimum bit error rate (BER) in turbulence channels with pointing errors. In Ref.[10], joint beam width and spatial coherence length optimization is proposed to maximize average capacity under the combined effects of atmospheric turbulence and pointing errors. In the aforementioned works, the bias errors are assumed to be zero, and the pointing errors follow a Rayleigh distribution. However, given that perfect calibration is not always achieved, bias errors could emerge. Thus, pointing errors follow a Rician distribution^[11-12].

To achieve the highest possible link reliability performance, the optimum beam divergence angle in the presence of pointing errors without bias errors and the optimum beam divergence angle with consideration of bias errors are investigated. Given BER and bit rate, link reliability is maximized through beam divergence angle selection in the presence of pointing errors.

2 System model and definition

2.1 Received signal model

To investigate the influence of pointing errors on intersatellite optical communication systems, we herein introduce a received signal model. We consider an intersatellite optical link (IOL) that employs on- off keying (OOK) modulation with intensity modulation/direct detection (IM/DD), which is widely used in practical optical communication systems. The transmitted data are modulated onto the instantaneous intensity of an optical beam, which is then transmitted with misalignment to the receiver; the received optical power is then converted to an electrical signal through a photodiode with responsivity R_v . The received signal y at the detector can then be modeled as^[7,12-13]

$$y = R_v A_r \eta h_p x + n , \qquad (1)$$

where $x \ge 0$, is the transmitted intensity signal with an average optical power $E\{x\} \le P_t$, P_t is the average transmitted optical power, $A_r = \pi D_r^2/4$ is the receiver antenna area, D_r is the receiver diameter, η represents the optical efficiency of the transmitter and receiver, n is a signal-independent zero-mean white Gaussian noise with variance σ_n^2 , and h_p denotes a random fluctuation of the optical intensity resulting from the pointing error of the intersatellite optical channels.

For a Gaussian beam of half-width divergence angle $\theta_{\rm b}$, the channel fading factor $h_{\rm p}$ resulting from the geometric spread with a pointing error angle $\theta_{\rm p}$ at communication distance *L* from the transmitter can be expressed as^[12-13]

$$h_{\rm p} = \frac{2}{\pi L^2 \theta_{\rm b}^2} \exp\left(-\frac{2\theta_{\rm p}^2}{\theta_{\rm b}^2}\right), \quad 0 \le h_{\rm p} \le \frac{2}{\pi L^2 \theta_{\rm b}^2} \quad . \tag{2}$$

2.2 Link reliability

System performance is conventionally described by BER, link reliability, and bit rate. In accordance with the aforementioned system model, the BER of the OOK modulation with IM/DD is given $by^{[7,14]}$

$$Q(h_{\rm p}) = \frac{1}{2} \operatorname{erfc}\left(\frac{R_{\rm v}A_{\rm r}\eta h_{\rm p}P_{\rm t}}{\sigma_{\rm v}2R_{\rm b}}\right),\tag{3}$$

where $\operatorname{erfc}(\cdot)$ is the complementary error function, R_b is the bit rate, and s is a coefficient related to noise standard deviation σ_{μ} ,

$$\sigma_{\rm p} = \sigma_{\rm q} \overline{R_{\rm p}} \,. \tag{4}$$

If Q_{T} is the target BER of a communication link and if

$$Q(h_{\rm pT}) = Q_{\rm T} , \qquad (5)$$

where $h_{_{\mathbb{P}^{T}}}$ is the threshold of the channel fading factor, then link reliability φ is the probability that the system BER is less than $Q_{_{T}}$ ^[7]. According to Eqs.(3) and (5),

$$\varphi = \operatorname{prob}[Q(h_p) < Q_T] = \operatorname{prob}[h_p > h_{pT}], \qquad (6)$$

where $prob(\cdot)$ denotes the probability of an event.

2.3 Pointing error model

The received optical intensity fluctuation, which is caused by pointing errors, has a significant impact on link reliability and bit rate. The probability density function (PDF) of radial pointing error angle θ_{p} can be modeled as a Rician distribution given by^[11-12]

$$f_{\theta_{p}}(\theta_{p}) = \frac{\theta_{p}}{\sigma_{p}^{2}} \exp\left(-\frac{\xi^{2} + \theta_{p}^{2}}{2\sigma_{p}^{2}}\right) I_{0}\left(\frac{\xi\theta_{p}}{\sigma_{p}^{2}}\right), \tag{7}$$

where σ_{p} is the standard deviation of the pointing error, ξ is the bias error angle, and $I_0(\cdot)$ is the modified 0th– order Bessel function of the first kind. The bias error is slowly time– varying and corresponds to a static portion while the other part of the pointing error is frequently time–varying and corresponds to a dynamic component known as a dynamic error. When the bias error is zero (i.e., $\xi=0$), the PDF of the radial pointing error angle θ_{p} becomes a Rayleigh distribution given by

$$f_{\theta_{p}}(\theta_{p}) = \frac{\theta_{p}}{\sigma_{p}^{2}} \exp\left(-\frac{\theta_{p}^{2}}{2\sigma_{p}^{2}}\right).$$
(8)

3 Link reliability maximization

3.1 Link reliability maximization with zero bias error

To improve link performance, we first investigate the link reliability and bit rate of the IOL in the presence of pointing errors without bias errors. With consideration of the PDF of θ_p without bias errors, as given by Eq.(8), and the relation between h_p and θ_p , as given by Eq.(2), the PDF of h_p can be expressed as

$$f_{h_{p}}(h_{p}) = \frac{\theta_{b}^{2}}{4\sigma_{p}^{2}} \cdot \left(\frac{\pi L^{2} \theta_{b}^{2}}{2}\right)^{\frac{\theta_{b}^{2}}{4\sigma_{p}^{2}}} \cdot h_{p}^{\frac{\theta_{b}^{2}}{4\sigma_{p}^{2}}-1}.$$
(9)

Combining Eqs. (6) and (9), we analytically obtain link reliability φ as

$$\varphi = \int_{h_{pT}}^{\infty} f_{h_{p}}(h_{p}) \mathrm{d}h_{p} = 1 - \left(\frac{\pi L^{2} \theta_{b}^{2}}{2} h_{pT}\right)^{\frac{\nu_{b}}{4\sigma_{p}^{2}}},$$
(10)

where h_{pT} can be obtained by combining Eqs. (3) and (5), as in

$$h_{\rm pT} = \frac{\sigma \sqrt{2R_{\rm b}}}{R_{\rm v}A_{\rm r}\eta P_{\rm t}} {\rm erfc}^{-1} (2Q_{\rm T}) \,. \tag{11}$$

According to Eqs. (10) and (11), the transmitted power P_{i} , distance L, and random pointing error σ_{p} are

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assumed to be constant. The variation of link reliability φ as a function of beam divergence angle θ_b at various desired bit rates (R_b) is shown in Fig.1. The parameters used in these numerical examples are shown in Table 1^[7,15]. In Fig.1, the existence of the maximum link reliability corresponding to the optimum beam divergence half angle θ_b^{opt} can be clearly observed in all four curves.



Fig.1 Link reliability versus half-width divergence angle $\theta_{\rm b}$ for various desired bit rates when bias error is zero Therefore, the optimum beam divergence angle $\theta_{\rm b}^{\rm opt}$ can be selected to maximize link reliability φ given target BER $Q_{\rm T}$, distance L, and bit rate $R_{\rm b}$. According to Eq.(10), $\theta_{\rm b}^{\rm opt}$ can be obtained theoretically and analytically using the following condition:

$$\frac{\partial \varphi}{\partial \theta_{\rm b}} = 0 , \qquad (12)$$

which yields

$$\theta_{\rm b}^{\rm opt} = \sqrt{\frac{2}{\exp(1) \cdot \pi L^2 h_{\rm pT}}} \,. \tag{13}$$

This expression is an analytic expression of optimum beam divergence angle $\theta_{\rm b}^{\rm opt}$ in the presence of pointing errors without bias errors.

Parameter	Symbol	Value	
Detector responsivity	$R_{ m v}$	0.5	
Optical efficiency	η	0.8	
Transmitted optical power	P_{t}	3 W	
Receiver diameter	$D_{ m r}$	250 mm	
Noise standard deviation (at 1 Gbit/s)	$\sigma_{ m n}$	$5 \times 10^{-7} \mathrm{A}$	
Distance	L	4000 km	
Required BER	$Q_{^{\mathrm{T}}}$	1×10^{-99}	
Pointing error standard deviation	$\sigma_{ m p}$	3 μrad	

Table 1 System parameters

3.2 Link reliability maximization with nonzero bias error

In practical systems, perfect calibration cannot always be achieved. This condition leads to the emergence of nonzero bias errors. Therefore, without loss of generality, we investigate the link reliability and bit rate of the IOL with consideration of bias errors. Combining Eq. (2) and the PDF of θ_p with the bias error given by Eq.(7), we can rewrite the PDF of h_p as^[12]

$$f_{h_{p}}(h_{p}) = \frac{\beta}{\alpha^{\beta}} \exp\left(-\frac{\xi^{2}}{2\sigma_{p}^{2}}\right) \cdot h_{p}^{\beta-1} \cdot I_{0}\left[\frac{\xi}{\sigma_{p}^{2}}\sqrt{-\frac{-\theta_{b}^{2}\ln(h_{p}/\alpha)}{2}}\right],$$
(14)

where $\alpha = 2/(\pi L^2 \theta_b^2)$, $\beta = \theta_b^2/(4\sigma_p^2)$, and $0 \le h_p \le \alpha$. Thus, according to Eqs.(6) and (14), we can obtain link reliability φ in the presence of pointing jitter with bias error as

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$$\varphi = \frac{\beta}{\alpha^{\beta}} \exp\left(-\frac{\xi^{2}}{2\sigma_{p}^{2}}\right) \int_{h_{pT}}^{\alpha} h_{p}^{\beta-1} \cdot I_{0}\left[\frac{\xi}{\sigma_{p}^{2}} \sqrt{\frac{-\theta_{b}^{2} \ln(h_{p}/\alpha)}{2}}\right] dh_{p}, \qquad (15)$$

Using a series representation of $I_0(\cdot)^{[16]}$ and letting $x = h_n/\alpha$, we obtain

$$\varphi = \beta \exp\left(\frac{-\xi^2}{2\sigma_p^2}\right) \sum_{m=0}^{\infty} \left[\frac{1}{(m!)^2} \left(-\frac{\xi^2 \theta_b^2}{8\sigma_p^4}\right)^m \int_{h_{pT}/\alpha}^1 x^{\beta^{-1}} (\ln x)^m dx\right].$$
(16)

Applying an integral identity ^[16] to Eq.(16), we obtain

$$\varphi = \beta \exp\left(-\frac{\xi^2}{2\sigma_p^2}\right) \sum_{m=0}^{\infty} \left\{ \frac{1}{\left(m!\right)^2} \left(-\beta \frac{\xi^2}{2\sigma_p^2}\right)^m \times \left[\frac{(-1)^m m!}{\beta^{m+1}} - \frac{1}{m+1} \left(\frac{h_{pT}}{\alpha}\right)^\beta \sum_{n=0}^m \frac{(-1)^n (m+1)!}{\beta^{n+1} (m-n)!} \left(\ln \frac{h_{pT}}{\alpha}\right)^{m-n}\right] \right\}.$$
(17)

For obtaining the maximum link reliability, the preceding equation cannot be solved analytically using the condition $d\varphi/d\theta_b=0$; nevertheless, numerical results can be obtained. Such results are shown in Fig.2. Link reliabilities φ are plotted as a function of θ_b/σ_p and desired bit rate R_b for the fixed transmitted power P_t , distance L, bias error ξ , and random pointing error σ_p . The parameters used in these numerical results are also shown in Table 1. Bias error ξ is assumed to be 10 µrad. The numerical calculation is performed with a relative tolerance of 10^{-12} . As shown in Fig.2, an optimum value of θ_b/σ_p maximizes link reliability, and for every R_b , the numerical results of the optimum ratio $(\theta_b/\sigma_p)^{\text{opt}}$ can be obtained. By applying the least squares fit to these calculated values, an analytic approximation for $(\theta_b/\sigma_p)^{\text{opt}}$ as a function of the desired bit rate R_b is given by

$$\left(\theta_{\rm b}/\sigma_{\rm p}\right)^{\rm opt} = a_{10}R_{\rm b}^{10} + a_{9}R_{\rm b}^{9} + a_{8}R_{\rm b}^{8} + a_{7}R_{\rm b}^{7} + a_{6}R_{\rm b}^{6} + a_{5}R_{\rm b}^{5} + a_{4}R_{\rm b}^{4} + a_{3}R_{\rm b}^{3} + a_{2}R_{\rm b}^{2} + a_{1}R_{\rm b}^{1} + a_{0}, \qquad (18)$$

where the coefficients a_i (*i*=0,1,2...,10) are as presented in Table 2. The accurate numerical results and the estimated values from the analytic approximation of Eq.(18) are plotted in Fig.3. The difference between the numerical and estimated results is less than 0.12% when 60.96 Mbit/s< R_b <1.5 Gbit/s.



Fig.2 Link reliability versus ratio of beam divergence angle to pointing error for various desired bit rates when bias error is nonzero



Fig.3 Comparison of the numerical results of optimum ratio $(\theta_b/\sigma_p)^{opt}$ and its estimated values from analytic approximation

Table 2	2 (Coefficients f	for appro	ximate exp	ression	describi	ing the	e optimum	beam	divergence an	gle f	for a	desired	bit	rate
							-			-	-				

Coefficient	Value	Coefficient	Value
a_{\circ}	15.2914	a_{6}	$1.2912{ imes}10^{-50}$
$a_{\scriptscriptstyle 1}$	-8.9748×10^{-8}	a_7	-9.8651×10^{-60}
a_2	5.7629×10^{-16}	a_{s}	$4.7466{ imes}10^{{ imes}69}$
$a_{\scriptscriptstyle 3}$	-2.3743×10^{-24}	a_9	-1.3032×10^{-78}
$a_{\scriptscriptstyle 4}$	6.3206×10^{-33}	$a_{\scriptscriptstyle 10}$	$1.5557{ imes}10^{-88}$
a_{5}	-1.1088×10^{-41}		

4 Conclusion

A link design method for intersatellite optical communication systems with misalignment fading is

presented. Link reliability is optimized through beam divergence angle selection with a target BER and bit rate. The analytic expression of the optimum beam divergence angle θ_{b}^{opt} in the presence of pointing errors with zero bias errors is obtained. The numerical results and analytic approximation of $(\theta_{b}/\sigma_{p})^{opt}$ in the presence of misalignment fading with nonzero bias errors are obtained. In some cases, such as when the bias error is much smaller than the dynamic error, the influence of the former can be ignored. In general, however, bias errors must be considered because they greatly affect system performance. The results of this work and the proposed link design method may be useful in the parametric estimation optimization of intersatellite optical communication systems.

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